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## GATE

Graduate Aptitude Test in Engineering


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## NETWORK

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## COINTENTS

## CHAPTER

PAGE

1. BASIC CONCEPTS ..... 1-43
2. NETWORK LAWS ..... 44-112
3. A.C.ANALYSIS ..... 113-169
4. MAGNETICALLY COUPLED CIRCUIT. ..... 170-189
5. NETWORK THEOREMS ..... 190-288
6. TRANSIENT ANALYSIS ..... 289-362
7. LAPLACE TRANFORMATION AND ITS APPLICATION IN ..... 363-379
8. RESONANCE ..... 380-421
9. TWO PORT NETWORKS ..... 422-507
10. GRAPH THEORY ..... 508-533
11. NETWORK FUNCTIONS ..... 534-577

## CHAPTER - 1

## BASIC CONCEPTS

### 1.1INTRODUCTION

### 1.1.1 Charge

Charge can be classified as:

1. Stationary Charge
2. Dynamic Charge

## 1. Stationary Charge

Stationary charge does not result into electric current because the flow of current means charge moving with net rate across any cross section.
(i) Any electric circuit should always follow law of conservation of charge and law of conservation of energy.
(ii) Circuit theory is analysed always at low frequency and field theory always at high frequency.
(iii) Transit time effect is always neglected at low frequency because
$T \gg t_{r}$
Where T is time period of sinosdual signal
$\mathrm{t}_{\mathrm{r}}$ is Transit Time (time taken by signal effect to travel from one point to another point).
(iv) Elemental law is obeyed only at low frequency such as ohm's law. It is not applicable at high frequency because of distributed nature of element.
(v) Elemental law always depend upon the nature of element


## For different Element, Different Form of Ohm's Law is present

(i) In time domain, the ohm's law are applicable and also in frequency domain.


Current flowing out of this body is given by equation of continuity as below
$\mathrm{I}=\oint \overrightarrow{\mathrm{J}} \cdot \overrightarrow{\mathrm{ds}}=-\frac{\mathrm{dQ}_{\text {in }}}{\mathrm{dt}}$
This equation gives the law of conservation of charge.
If $\frac{\mathrm{dQ}_{\text {in }}}{\mathrm{dt}}=0$; means no rate of change of charge within body then eq.(i) become


Example 1. The waveform of the current through an inductor of 10 H is shown in Fig. Sketch the waveform of the voltage across the inductor


## Solution.

The voltage across an inductor is

$$
\mathrm{v}_{\mathrm{L}}=\mathrm{L} \frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}
$$

(a) Between 0 and 0.1 s (that is, for $0<\mathrm{t}<0.1$ ), the derivative, $\frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}$, which is the slope of the $\mathrm{i}_{\mathrm{L}}$ curve, is constant since $\mathrm{i}_{\mathrm{L}}(\mathrm{t})$ is linear $\frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=$ slope of line OA
$=\frac{\mathrm{OA}}{\mathrm{AM}}=\frac{10}{0.1}=100 \mathrm{~A} / \mathrm{s}$ (constant)
$\therefore \mathrm{V}_{\mathrm{L}}=\mathrm{L} \frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=10 \times 100=1000 \mathrm{~V}$ (constant) for
$0<$ t $<0.1$
This constant value is plotted as a horizontal line $a b$ in fig.
(b) For $0.1<\mathrm{t}<0.2 \mathrm{~s}$ the current curve AB is horizontal, that is, its slope is zero
$\therefore \mathrm{v}_{\mathrm{L}}=\mathrm{L} \frac{\mathrm{di}}{\mathrm{dt}}=10 \times 0=0 \mathrm{~V}$
For $0.1<\mathrm{t}<0.2 \mathrm{~s}$
This is plotted as horizontal line cd in fig.
(c) For $0.2<\mathrm{t}<0.4 \mathrm{~s}$, the current curve is BC. The slope of $B C$ is negative.
$\therefore \frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=$ slope of line BC
$=\frac{\mathrm{BN}}{\mathrm{NC}}=\frac{10}{0.2}=-50 \mathrm{~A} / \mathrm{s}$
$\therefore v_{\mathrm{L}}=\mathrm{L} \frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=10 \times 0=0$ for $0.4<\mathrm{t}<0.6 \mathrm{~s}$
The voltage waveform in the range $0.4<\mathrm{t}<$ 0.6 s is by fh in Fig. (b)
(e) For $0.6<\mathrm{t}<0.8 \mathrm{~s}$, the current curve is DE. I has a negative slope given by
$\frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=-\frac{\mathrm{EP}}{\mathrm{PD}}=-\frac{10}{0.2}=-50 \mathrm{~A} / \mathrm{s}$
$\therefore v_{\mathrm{L}}=\mathrm{L} \frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=10 \times(-50)=-500 \mathrm{~V}$ for $0.6<\mathrm{t}$ $<0.8$ s
The voltage waveform in this range is show by km in fig.
(f) For $0.8<\mathrm{t}<0.9 \mathrm{~s}$, since the current curve EF is horizontal, $\frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=0$

$$
\therefore \mathrm{v}_{\mathrm{L}}=\mathrm{L} \frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=10 \times 0=0 \mathrm{~V}
$$

This is shown by curve $\ell \mathrm{n}$ in fig. 9 b )
(g) for $0.9<\mathrm{t}<1.0 \mathrm{~s}$, the current curve FG has a slope
$\frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=\frac{\mathrm{FQ}}{\mathrm{QG}}=\frac{10}{0.1}=100 \mathrm{~A} / \mathrm{s}$
$\therefore v_{\mathrm{L}}=\mathrm{L} \frac{\mathrm{di}_{\mathrm{L}}}{\mathrm{dt}}=10 \times 100=1000 \mathrm{~V}$
This is shown by curve pq in Fig.(b)
Overall voltage waveform is shown below:



1. $R_{A}$ and $R_{B}$ are the input resistances of circuits as shown below. The circuits extend infinitely in the direction shown. Which one of the statements is TRUE?

(a) $R_{A}=R_{B}$
(b) $\mathrm{R}_{\mathrm{A}}=\mathrm{R}_{\mathrm{B}}=0$
(c) $R_{A}<R_{B}$
(d) $R_{B}=R_{A} /\left(1+R_{A}\right)$
2. In the circuit shown in the figure, the magnitude of the current (in amperes) through $\mathrm{R}_{2}$ is $\qquad$

[GATE - 2016]
3. An incandescent lamp is marked $40 \mathrm{~W}, 240 \mathrm{~V}$. If resistance at room temperature $\left(26^{\circ} \mathrm{C}\right)$ is $120 \Omega$, and temperature coefficient of resistance is $4.8 \times 10^{-3} /{ }^{\circ} \mathrm{C}$, then its ' ON ' state filament temperature in ${ }^{\circ} \mathrm{C}$ is approximately
[GATE - 2014]
4. The circuit shown in the figure represents a

[GATE - 2014]
(b) Voltage controlled current source
(c) Current controlled current source
(d) Current controlled voltage source
5. A fully charged mobile phone with a 12 V battery is good for a 10 minute talk-time. Assume that during the talk - time the battery delivers a constant current of 2 A and its voltage drops linearly from 12 V to 10 V as shown in the figure. How much energy does the battery deliver during the talk - time?

[GATE - 2009]
(a) 220 J
(b) 12 kJ
(c) 13.2 kJ
(d) 14.4 J

Common data for Q. 6 \& Q. 7
The current $\mathrm{i}(\mathrm{t})$ sketched in the figure flows through a initially uncharged 0.3 nF capacitor.

6. The charge stored in the capacitor at $\mathrm{t}=5 \mu \mathrm{~s}$, will be
[GATE - 2008]
(a) 8 nC
(b) 10 nC
(c) 13 nC
(d) 16 nC


1. A sinusoidal voltage waveform has frequency 50 Hz and RMS voltage 30 V . The equation representing the waveform is
[EC ESE - 2018]
(a) $V=30 \sin 50 t$
(b) $V=60 \sin 20 t$
(c) $V=42.42 \sin 314 t$
(d) $V=84.84 \sin 314 t$
2. A waveform, shown in the figure is applied to a resistor of $20 \Omega$. The power dissipated in the resistor is

[EC ESE - 2018]
(a) 100 W
(b) 600 W
(c) 900 W
(d) 1000 W
3. The Kirchhoff's current law works on the principle of conservation of
4. Charge
5. Energy
6. Power

Which of the above is/are correct?
[EC ESE - 2018]
(a) 1 only
(b) 2 only
(c) 3 only
(d) 1, 2 and 3
4. The voltage and current waveforms for an element are shown in the figure.



The circuit element and its value are
[EE ESE - 2017]
(a) Capacitor, 2 F
(b) Inductor, 2 H
(c) Capacitor, 0.5 F
(d) Inductor, 0.5 H
5. The equivalent resistance between the points $A$ and $D$ is

(a) $10 \Omega$
(b) $20 \Omega$
(c) $30 \Omega$
(d) $40 \Omega$
6. The capacitance of each capacitor is $\mathrm{C}=3 \mu \mathrm{~F}$ in the figure shown. The effective capacitance between point A and B is

[EC ESE - 2015]
(a) $2 \mu \mathrm{~F}$
(b) $3 \mu \mathrm{~F}$
(c) $4 \mu \mathrm{~F}$
(d) $5 \mu \mathrm{~F}$
7. A network N consists of resistors, dependent and independent voltage and current sources. If the current in one particular resistance is 1 A , it will be doubled if the values of all the
[EC ESE - 2014]
(a) Independent voltage sources are doubled.
(b) Independent current sources are doubled.
(c) Dependent and independent voltage and current sources are doubled
(d) Independent voltage and current sources are doubled
8. A voltage of 2000 V exists across 1 cm insulating space between two parallel conducting plates. An electron of charge 1.6 $\times 10^{-19}$ coulomb is introduced into the space. The force on the electron is
[EC ESE - 2014]
(a) $18.2 \times 10^{-26} \mathrm{~N}$
(b) $3.2 \times 10^{-14} \mathrm{~N}$
(c) $1.6 \times 10^{-19} \mathrm{~N}$
(d) $4.5 \times 10^{26} \mathrm{~N}$

## CHAPTER - 2

NETWORK LAWS

### 2.1 KIRCHOFF'S VOLTAGE LAW (KVL)

It states that algebraic sum of all voltages in a closed path or loop is zero

$$
\sum_{\text {loop }} \mathrm{V}=0
$$

For writing KVL start from any point in the loop and come to the same point via transversing the path of closed loop. While doing so take voltage rises as positive and voltage drops as negative then
$\Sigma$ voltage rise $+\Sigma$ voltage drops $=0$
$\mathrm{Q}^{+}$positive is moving
A-B is energy is gained say $W_{1}$
B-C is energy lost say $W_{2}$
C-D is energy lost say $W_{3}$
D-A is energy lost say $\mathrm{W}_{4}$
by consecration of energy
$\mathrm{W}_{1}=\mathrm{W}_{2}+\mathrm{W}_{3}+\mathrm{W}_{4}$ or
$\mathrm{W}_{1}-\mathrm{W}_{2}-\mathrm{W}_{3}-\mathrm{W}_{4}=0$
Divide by $\mathrm{Q}=\frac{\mathrm{W}_{1}}{\mathrm{Q}}-\frac{\mathrm{W}_{2}}{\mathrm{Q}}-\frac{\mathrm{W}_{3}}{\mathrm{Q}}-\frac{\mathrm{W}_{4}}{\mathrm{Q}}=0$
$\Rightarrow V_{1}-V_{2}-V_{3}-V_{4}=0$
$\Sigma \mathrm{V}=0$ (In a loop )
Example. KVL in this loop starting from a in clockwise direction is
$-V_{7}-V_{8}-V_{5}+V_{6}=0$
$\Rightarrow V_{6}=V_{7}+V_{8}+V_{5}$


The basis of the law is that if we start from a particular junction and go round the mesh till we come back to the starting point, then we must be at the same potential with which we started. Hence it means that all the sources of e.m.f. met on the way must necessarily be equal to the voltage drops in the resistances, every voltage being given its proper sign, plus or minus.

## (1)

## Example 1.



Find equivalent resistance across AB

## Solution.

Arbitrary assigning $\mathrm{V}_{1}, \mathrm{~V}_{2} \& \mathrm{~V}_{3}$ etc.


Potential difference (p.d.) across $2 \Omega$ resistance $=\mathrm{V}_{1}-\mathrm{V}_{2}$
Potential difference (p.d.) across $6 \Omega$ resistance $=V_{2}-V_{1}$
Potential difference (p.d.) across $3 \Omega$ resistance $=V_{1}-V_{2}$
We will not consider sign of p.d. only magnitude is of concern, now because all the resistances have same p.d., they all are in parallel hence
$\frac{1}{\mathrm{R}_{\mathrm{AB}}}=\frac{1}{2}+\frac{1}{6}+\frac{1}{3} \Rightarrow \mathrm{R}_{\mathrm{AB}}=1 \Omega$

Example 2. Find $\mathrm{R}_{\mathrm{AB}}$.
Solution.

$$
\begin{aligned}
& \Rightarrow \mathrm{R}_{\mathrm{AB}}=\frac{\mathrm{R}}{4}+\frac{\mathrm{R}}{4}+\frac{\mathrm{R}}{2}+\frac{\mathrm{R}}{2}=\frac{3 \mathrm{R}}{2}
\end{aligned}
$$

## Example 3.



Find equivalent resistance between AB

Solution.
All resistance are $\mathrm{R} \Omega$ find $\mathrm{R}_{\mathrm{AB}}$


Now here AB is line of symmetry point $2 \& 8$ can be joined point $3,9,7$ can be joined and point 2 \& 6 can be joined.
The circuit can be redrawn as


In this problem $A B$ is not line of symmetry however perpendicular bisector of $A B$ that is $C D$ is a line of symmetry. Hence points C, E \& $D$ are at same potentials hence $R_{C E} \& R_{E D}$ can be removed from the circuit.

## Example 4.



Find $\mathrm{R}_{\mathrm{AB}}$ (equivalent resistance across diagonal of a cube)
Solution.
For solving such problems. Assume current I is entering at A then this i will come out of B then this current is distributed as per symmetry

1.Consider the circuit graph shown in fig. Each branch of circuit graph represent a circuit element. The value of voltage $v_{1}$ is

(a) -30 V
(b) 25 V
(c) -20 V
(d) 15 V
2. In the circuit of figure, power absorbed by the unknown elements are given. Then the current $\mathrm{I}_{\mathrm{x}}$ in the circuit is

(a) 1.5 A
(b) 6.0 A
(c) 3.0 A
(d) None
3. In the circuit of figure, the ideal switch $S$ alternatively stays at position $A$ for 1 ms and at position $B$ for 4 ms . The average value of current $i(t)$ is

(a) 0.6 mA
(b) 1.5 mA
(c) 0.75 mA
(d) 2.5 mA
4.If a resistor of $10 \Omega$ is placed in parallel with voltage source in the circuit of figure, the current $i$ will be

(a) Increased
(b) Decreased
(c) Constant
(d) It is not possible to say
5.Two elements are connected in series as shown in figure. Element-1 supplies 36W of power. Element -2
(a) Absorbs 72 W
(b) Absorbs 36 W
(c) Supplies 72W
(d) Absorb 144 W
6. The voltage $\mathrm{v}_{\mathrm{L}}$ in figure is related to the current $\mathrm{i}_{\mathrm{L}}$ according to

$$
\mathrm{v}_{\mathrm{L}}=\left\{\begin{array}{cc}
16-4 \mathrm{i}_{\mathrm{L}}^{2} & 0 \leq \mathrm{i}_{\mathrm{L}} \leq 2 \\
0 & \mathrm{i}_{\mathrm{L}}>2
\end{array}\right.
$$



The value of $i_{L}$ that maximizes the power absorbed by the load is
(a) 0 A
(b) 1.333 A
(c) 2 A
(d) 1.155 A
7. In the circuit shown in figure, the value of voltage $v_{0}$ is

## ASSIGNMENT

1. The figure shows a dependent current source.


It
(a) Absorbs 80 W
(b) Delivers 80 W
(c) Absorbs 100 w
(d) Delivers 100W
2. In the current shown ,the power supplied by the voltage source is zero


Current $i_{1}$ is
(a) 2 A
(b) 3 A
(c) 1.6 A
(d) 4 A
3. Following network shown is the star network and its corresponding delta network.

$\mathrm{R}_{12}, \mathrm{R}_{23}$ and $\mathrm{R}_{31}$ are respectively
(a) $1.6 \Omega, 2.1 \Omega, 2.5 \Omega$
(b) $2.5 \Omega, 5 \Omega, 6 \Omega$
(c) $2 \Omega, 3 \Omega, 4 \Omega$
(d) $2 \Omega, 3 \Omega, 5 \Omega$
4. The rms value of the voltage function $\mathrm{V}=50+100 \sin \omega \mathrm{t}+50 \sin 3 \omega \mathrm{t}$ is
(a) 91 V
(b) 93.5 V
(c) 102.6 V
(d) 98.3 V
5. The input resistance of the circuit shown below is

(a) $6 \Omega$
(b) $1.95 \Omega$
(c) $2.24 \Omega$
(d) $3.36 \Omega$
6. $\mathrm{R}_{\text {in }}$ for the network shown is


Each R is $10 \Omega$
(a) $10 \Omega$
(b) $15 \Omega$
(c) $3 \Omega$
(d) $11.43 \Omega$
7. The voltage $\mathrm{V}_{\mathrm{A} 0}$ in the circuit is

(a) -0.4 V
(b) +0.4 V
(c) 1.6 V
(d) 12 V


1. The equivalent resistance seen between the terminals $(a, b)$ is:

(a) $2 \Omega$
(b) $4 \Omega$
(c) $1 \Omega$
(d) Not Possible
2. All the resistances in the figure are $1 \Omega$ each. The value of I will be:

(a) $1 / 15 \mathrm{~A}$
(b) $2 / 15 \mathrm{~A}$
(c) $4 / 15 \mathrm{~A}$
(d) $8 / 15 \mathrm{~A}$
3. In the circuit below, $\mathrm{V}_{1}=40 \mathrm{~V}$ when R is $10 \Omega$, when R is zero, the value of $\mathrm{V}_{2}$ will be:

(a) 40 V
(b) 30 V
(c) 20 V
(d) 10 V
4. In the circuit shown below, the voltage across $2 \Omega$ resistor is 20 V . The $5 \Omega$ resistor connected between the terminals A and B can be replaced by an ideal.

(a)Voltage source of 25 V with + ve terminal upward
(b)Voltage source of 25 V with +ve terminal downward
(c)Current source of 2A upward.
(d)Current source of 2A downward.
5. Consider the following circuit:


What is the power delivered to resistor in the above circuit?
(a) -15 W
(b) 0 W
(c) 15 W
(d) Cannot be determined unless the value of R is known.
6. In the circuit shown in the figure, the power dissipated in $30 \Omega$ resistor will be maximum if the value of R is:

(a) $30 \Omega$
(b) $16 \Omega$
(c) $9 \Omega$
(d) Zero
7. In the circuit shown in the figure, for $\mathrm{R}=$ $20 \Omega$, the circuit ' I ' is 2 A . When R is $10 \Omega$, the current 'I' would be:

8. In the figure below, the current of 1 A flows through the resistance of:

## GATE QUESTIONS

1. Consider the network shown below with $\mathrm{R}_{1}=$ $1 \Omega, \mathrm{R}_{2}=2 \Omega$ and $\mathrm{R}_{3}=3 \Omega$. The network is connected to a constant voltage source of 11 V .


The magnitude of the current (in amperes, accurate to two decimal places) through the source is $\qquad$
[GATE - 2018]
2. A circuit consisting of dependent and independent sources is shown in the figure. If the voltage at Node -1 is -1 V , then the voltage at Node - 2 is $\qquad$ V.

[GATE - 2017]
3. In the circuit shown below, the voltage and current sources are ideal. The voltage $\left(\mathrm{V}_{\text {out }}\right)$ across the current source, in volts, is

(a) 0
(b) 5
(d) 20
4. In the circuit shown below, the node voltage $\mathrm{V}_{\mathrm{A}}$ is $\qquad$ V.

[GATE - 2016]
5. In the given circuit, the current supplied by the battery, in ampere, is $\qquad$ -.

[GATE - 2016]
6. In the figure shown, the current i (in ampere) is $\qquad$ -.

[GATE - 2016]
7. In the given circuit, each resistor has a value equal to $1 \Omega$


## 1. Statement I:

Two ideal current sources with currents $\mathrm{I}_{1}$ and $\mathrm{I}_{2}$ cannot be connected in parallel.

## Statement II:

Superposition theorem cannot be applied to ideal current sources if these sources are connected in cascade.
[EC ESE - 2017]

## Codes:

(a) Both Statement I and II are individually true and Statement II is the correct explanation of Statement I.
(b) Both Statement I and Statement II are individually true but Statement II is not the correct explanation of Statement I.
(c) Statement I is true but Statement II is false.
(d) Statement I is false but Statement II is true.
2. What is the current through the $8 \Omega$ resistance connected across terminals, M and N in the circuit?

[EE ESE - 2017]
(a) 0.34 A from M to N
(b) 0.29 A from M to N
(c) 0.29 A from n to M
(d) 0.34 A from N to M
3. Consider the following statements:

1. Flaming's rule is used where induced e.m.f is due to flux cutting.
2. Leng' $z$ law is used when the induced e.m.f is due to change in flux linkages.
3. Lenz's law is direct consequence of the law of conservation of energy.
Which of the above statements are correct?
[EC ESE - 2017]
(a) 1 and 2 only
(b) 1 and 3 only
(c) 2 and 3 only
(d) 1,2 and 3
4. Consider the following factors:
5. Number of turns of the coil
6. Length of the coil
7. Rea of cross-section of the coil 4. permeability of the core

On which of the above factors does inductance, depend?
[EC ESE - 2017]
(a) 1, 2 and 3 only
(b) 1, 3 and 4 only
(c) 1, 2, 3, and 4
(d) 2 and 4 only
5. Consider a packet switched network based on a virtual circuit mode of switching. The delay jutter for the packets of a session from the source node to the destination node is/are

1. Always zero
2. Non-zero
3. For some networks, zero

Select the correct answer using the code given below.
[EC ESE - 2017]
(a) 1
(b) 2 only
(c) 3 only
(d) 2 and 3
6. A network in which all the elements are physically separable is called a
[EC ESE - 2017]
(a) Distributed network
(b) Lumped network
(c) Passive network
(d) Reactive network
7. For the active network shown in figure, the value of $V / I$ is

### 3.1 AC THROUGH PURE OHMIC RESISTANCE ALONE


$\mathrm{V}=\mathrm{V}_{\mathrm{m}} \sin \omega \mathrm{t}$
$\mathrm{v}=\mathrm{i} \mathrm{R}$
$i=\frac{V_{m}}{R} \sin \omega t$
Current is max when $\sin w t=1$
i.e. $I_{m}=\frac{V_{m}}{R}$
$\therefore \mathrm{i}=\mathrm{I}_{\mathrm{m}} \sin \omega \mathrm{t}$

### 3.2 AC THROUGH PURE INDUCTANCE ALONE

Whenever an alternating voltage is applied to a purely inductive coil, a back emf is produced due to the self-inductance of the coil

$\mathrm{v}=\mathrm{L} \frac{\mathrm{di}}{\mathrm{dt}}$
$\mathrm{V}_{\mathrm{m}} \sin \omega \mathrm{t}=\frac{\mathrm{Ldi}}{\mathrm{dt}} \Rightarrow \mathrm{di}=\frac{\mathrm{V}_{\mathrm{m}}}{\mathrm{L}} \sin \omega \mathrm{tdt}$

$\mathrm{i}=\frac{\mathrm{V}_{\mathrm{m}}}{\mathrm{L}} \int \sin \omega \mathrm{tdt}$

## 9 <br> WORKBOOK

Example 1. Average and RMS value of full sine wave

$\mathrm{v}(\mathrm{t})=\mathrm{V}_{\mathrm{m}} \sin \omega \mathrm{t}$
$\mathrm{v}_{\mathrm{av}}=\frac{\mathrm{v}_{\mathrm{m}}}{2 \pi} \int_{0}^{2 \pi} \mathrm{v}_{\mathrm{m}} \sin \omega \mathrm{t} \mathrm{d}(\omega \mathrm{t})=0$
Average value of full cycle of sine and cosine is zero.
Average value for half cycle
$\mathrm{v}_{\mathrm{av}}=\frac{\mathrm{v}_{\mathrm{m}}}{\pi} \int_{0}^{\pi} \mathrm{v}_{\mathrm{m}} \sin \omega \mathrm{t} \mathrm{d}(\omega \mathrm{t})$
$=\frac{\mathrm{V}_{\mathrm{m}}}{\pi}[-\cos \omega \mathrm{t}]_{0}^{\pi}=\frac{\mathrm{V}_{\mathrm{m}}}{\pi}[-\cos \pi+\cos 0]=\frac{2 \mathrm{~V}_{\mathrm{m}}}{\pi}$
RMS value
$\mathrm{V}_{\mathrm{rms}}=\sqrt{\frac{1}{\pi} \int_{0}^{\pi} \mathrm{V}_{\mathrm{m}}^{2} \sin ^{2} \omega \mathrm{t} \mathrm{d}(\omega \mathrm{t})}$
$=\mathrm{V}_{\mathrm{m}} \sqrt{\frac{1}{\pi} \int_{0}^{\pi} \frac{1}{2}(1-\cos 2 \omega \mathrm{t}) \mathrm{d}(\omega \mathrm{t})}$
$=\frac{\mathrm{V}_{\mathrm{m}}}{\sqrt{2}}$
Example 2. Find average and RMS value of half wave rectified sine wave

## Solution.



Average value
$\mathrm{V}_{\mathrm{av}}=\frac{1}{2 \pi} \int_{0}^{\pi} \mathrm{V}_{\mathrm{m}} \sin \omega \mathrm{t} \mathrm{d}(\omega \mathrm{t})=\frac{\mathrm{V}_{\mathrm{m}}}{\pi}$
$\mathrm{V}_{\mathrm{rms}}=\sqrt{\frac{1}{2 \pi} \int_{0}^{\pi} \mathrm{V}_{\mathrm{m}}^{2} \sin ^{2} \omega \mathrm{td}(\omega \mathrm{t})}=\frac{\mathrm{V}_{\mathrm{m}}}{2}$

Example 3. Find out average and RMS value of saw tooth wave form

## Solution.


$\mathrm{v}(\mathrm{t})=\frac{\mathrm{V}_{\mathrm{m}}}{\mathrm{T}} \times \mathrm{t}$ for $0<\mathrm{t}<\mathrm{T}$
Average value
$\mathrm{V}_{\mathrm{av}}=\frac{1}{\mathrm{~T}} \int_{0}^{\mathrm{T}} \frac{\mathrm{V}_{\mathrm{m}}}{\mathrm{T}} \mathrm{tdt}$
$=\left.\frac{\mathrm{V}_{\mathrm{m}}}{\mathrm{T}^{2}} \frac{\mathrm{t}^{2}}{2}\right|_{0} ^{\mathrm{T}}=\frac{\mathrm{V}_{\mathrm{m}}}{2}$
RMS value $V_{r m s}=\sqrt{\frac{1}{T} \int_{0}^{T} \frac{V_{m}^{2}}{T^{2}} t^{2} d t}$
$=\sqrt{\frac{V_{m}^{2}}{T^{3}} \int_{0}^{T} t^{2} d t}$
$=\sqrt{\left.\frac{V_{m}^{2}}{T^{3}} \frac{t^{3}}{3}\right|_{0} ^{T}}=\frac{V_{m}}{\sqrt{3}}$
Example 4. Given $\mathrm{i}_{1}(\mathrm{t})=4 \cos \left(\omega \mathrm{t}+30^{\circ}\right) \mathrm{A}$ and $\mathrm{i}_{2}(\mathrm{t})=5 \sin \left(\omega \mathrm{t}-20^{\circ}\right) \mathrm{A}$, find their sum.
Solution.
$\mathrm{I}_{1}=4 \angle 30^{\circ}$
$\mathrm{i}_{2}=5 \cos \left(\omega \mathrm{t}-20^{\circ}-90^{\circ}\right)$
$=5 \cos \left(\omega t-110^{\circ}\right)$ And its phasor is
$\mathrm{I}_{2}=5 \angle-110^{\circ}$
$\mathrm{I}=\mathrm{I}_{1}+\mathrm{I}_{2}$
$=4 \angle 30^{\circ}+5 \angle-110^{\circ}$
$=3.464+2 \mathrm{j}-1.71-\mathrm{j} 4.698$
$=1.754-\mathrm{j} 2.698$
$=3.218 \angle-56.97^{\circ} \mathrm{A}$
Transforming this to time domain, we get

## ASSIGNMENT

1. The current flowing through a circuitcontaining passive elements like $\mathrm{R}, \mathrm{L}$ or C is $\mathrm{I}=15.5 \sin \left(2500 \mathrm{t}-145^{\circ}\right)$ with voltage source $\mathrm{V}=311 \sin \left(2500 \mathrm{t}+170^{\circ}\right)$. The impedance Z is
(a) $20 \angle-25^{\circ} \Omega$
(b) $20 \angle 25^{\circ} \Omega$
(c) $20+\mathrm{j} 20 \Omega$
(d) $14.14-\mathrm{j} 14.14 \Omega$
2. There is a pure element in series with $\mathrm{R}=$ $25 \Omega$ which causes the current to lag the voltage by $20^{\circ}$. It is $\qquad$ if frequency is 400 Hz .
(a) $35 \mu \mathrm{~F}$ capacitor
(b) 3.6 mH inductor
(c) $25 \mu \mathrm{~F}$ capacitor
(d) 2 mH inductor
3. There is a pure element in series with $\mathrm{R}=$ $50 \Omega$ which causes the current to lead the voltage by $30^{\circ}$. It is $\qquad$ if frequency is 500
Hz.
(a) $11 \mu \mathrm{~F}$ capacitor
(b) $22 \mu \mathrm{~F}$ capacitor
(c) 1.2 mH
(d) 2.4 mH inductor
4. In the $A C$ network shown in the figure. The phasor voltage $\mathrm{V}_{\mathrm{AB}}$ (in volts) is

(a) $7 \angle 30^{\circ}$
(b) $6 \angle 30^{\circ}$
(c) $21 \angle 30^{\circ}$
(d) $11 \angle 30^{\circ}$
5. The differential equation for the current $i(t)$ in the circuit shown below is

(a) $2 \frac{\mathrm{~d}^{2} \mathrm{i}}{\mathrm{dt}^{2}}+3 \frac{\mathrm{di}}{\mathrm{dt}}+\mathrm{i}(\mathrm{t})=-\sin \mathrm{t}$
6. The value of $L$ in the given circuit to keep power factor zero is
(b) $2 \frac{\mathrm{~d}^{2} \mathrm{i}}{\mathrm{dt}^{2}}+3 \frac{\mathrm{di}}{\mathrm{dt}}+\mathrm{i}(\mathrm{t})=\sin \mathrm{t}$
(c) $2 \frac{\mathrm{~d}^{2} \mathrm{i}}{\mathrm{dt}^{2}}+3 \frac{\mathrm{di}}{\mathrm{dt}}+\mathrm{i}(\mathrm{t})=\cos \mathrm{t}$
(d) $2 \frac{\mathrm{~d}^{2} \mathrm{i}}{\mathrm{dt}^{2}}+3 \frac{\mathrm{di}}{\mathrm{dt}}+\mathrm{i}(\mathrm{t})=-\cos \mathrm{t}$

## Common Data for Q. 6 \& Q. 7

Given below is the current and applied voltage in a series connection of two pure circuit elements.
$V=150 \sin \left(314 t+10^{\circ}\right)$ volts
$\mathrm{i}=15 \sin \left(314 \mathrm{t}-53.4^{\circ}\right)$ amperes
6. The circuit contains a
(a) Resistance of $5 \Omega$ and capacitor of $1 \mu \mathrm{~F}$
(b) Resistance of $4.47 \Omega$ and inductor of 0.028 H
(c) Resistance of $6.7 \Omega$ and capacitor of $0.01 \mu \mathrm{~F}$
(d) Resistance of $10 \Omega$ and inductor of 0.021 H
7. Consider the circuit shown in the figure


In the circuit $4 \mathrm{R}^{2} \mathrm{C}=3 \mathrm{~L}$, the resonance frequency $\omega_{0}$ is
(a) $\sqrt{\mathrm{LC}}$
(b) $\frac{1}{\sqrt{\mathrm{LC}}}$
(c) $\frac{1}{2 \pi \sqrt{\mathrm{LC}}}$
(d) $\frac{1}{2 \sqrt{\mathrm{LC}}}$
8. A voltage V is represented as $\mathrm{V}=50 \sin (\omega \mathrm{t}+$ $\left.30^{\circ}\right)-25 \sin \left(3 \omega \mathrm{t}-60^{\circ}\right)+16 \sin \left(5 \omega \mathrm{t}+45^{\circ}\right) \mathrm{V}$ R.M.S. value of the voltage is
(a) 41 volt
(b) 41.11 volt
(c) 91 volt
(d) 58.14 volt


1. A series R-L-C circuit is excited with a 50 v , 50 Hz sinusoidal source. The voltages across the resistance and the capacitance are shown in the figure. The voltage across the inductor $\left(\mathrm{V}_{\mathrm{L}}\right)$ is
$\qquad$ V.

[GATE - 2017]
2. A connection is made consisting of resistance A is series with a parallel combination of resistance B and C . Three resistors of value $10 \Omega, 5 \Omega, 2 \Omega$ are provided. Consider all possible permutations of the given resistors into the positions $\mathrm{A}, \mathrm{B}, \mathrm{C}$ and identify the configuration with maximum possible overall resistance; and also the ones with minimum possible overall resistance; and also the ones with minimum possible overall resistance. The ratio of maximum to minimum values of the resistances (up to second decimal place) is
[GATE - 2017]
3. The figure shows an RLC circuit excited by the sinusoidal voltage $100 \cos (3 \mathrm{t})$ Volts, where t is in seconds. The ratio $\frac{\text { amplitude of } \mathrm{V}_{2}}{\text { amplitude of } \mathrm{V}_{1}}$ is

4. In the circuit shown, the positive angular frequency $\omega$ (in radians per second ) at which
the magnitude of the phase difference between the voltages $V_{1}$ and $V_{2}$ equals $\frac{\pi}{4}$ radians, is $\qquad$ .

[GATE - 2017]
5. The mean square value of the given periodic waveform $f(t)$ is $\qquad$

[GATE - 2017]
6. The following figure shows the connection of an ideal transformer with primary to secondary turns ratio of $1: 100$. The applied primary voltage is 100 V (rms), $50 \mathrm{~Hz}, \mathrm{AC}$. The rms value of the current $I$, in ampere, is $\qquad$ -.

[GATE - 2016]
7. A resistance and a coil are connected in series and supplied from a single phase, 100 V , 50 Hz ac source as shown in the figure below. The rms values of plausible voltages across the resistance $\left(\mathrm{V}_{\mathrm{R}}\right)$ and coil $\left(\mathrm{V}_{\mathrm{C}}\right)$ respectively, in volts, are

8. A two-element series circuit is connected across an AC source given by
$\mathrm{e}=200 \sqrt{2} \sin (314 \mathrm{t}+20) \mathrm{V}$. The current is then found to be $i=10 \sqrt{2} \cos (314 t-25) V$. The parameters of the circuit are
[EE ESE - 2017]
(a) $\mathrm{R}=20 \Omega$ and $\mathrm{C}=160 \mu \mathrm{~F}$
(b) $\mathrm{R}=14.14 \Omega$ and $\mathrm{C}=225 \mu \mathrm{~F}$
(c) $\mathrm{L}=45 \mathrm{mH}$ and $\mathrm{C}=225 \mu \mathrm{~F}$
(d) $\mathrm{L}=45 \mathrm{mH}$ and $\mathrm{C}=160 \mu \mathrm{~F}$
9. Two resistors of $5 \Omega$ and $10 \Omega$ and an inductor L are connected in series across a $50 \cos \omega t$ voltage source. If the power consumed by the $5 \Omega$ resistor is 10 W , the power factor of the circuit is
[EE ESE - 2017]
(a) 1.0
(b) 0.8
(c) 0.6
(d) 0.4
10. Statement (I): One series RC circuit and the other series RL circuit are connected in parallel across at ac supply. The circuit exhibits two reasonance when L is variable.
Statement (II): The circuit has two values of 1 for which the imaginary part of the input admittance of the circuit is zero.
[EE ESE - 2017]
(a)Both Statement (I) and Statement (II) are individually true and Statement (II) is the correct explanation of Statement (I)
(b)Both Statement (I) and Statement (II) are individually true but Statement (II) is not the correct explanation of Statement (I)
(c)Statement (I) is true but Statement (II) is false.
(d)Statement (I) is false but Statement (II) is true
11. One of the basic characteristics of any steady state sinusoidal response of a linear R-L-C circuit with constant $\mathrm{R}, \mathrm{L}$ and C values is
[EC ESE - 2017]
(a)The output remains sinusoidal with its frequency being the same as that of the source
(b)The output remains sinusoidal with its frequency differing from that of the source
(c)The output amplitude equals the soruce amplitude
(d)The phase angle difference between the source and the output is always zero.
12. Three identical impedances are first connected in delta across a 3-phase balanced supply. If the same impedances are now connected in star across the same supply, then
[EC ESE - 2017]
(a)The phase current will be one - third
(b)The line current will be one - third
(c)The power consumed will be one -third
(d)The power consumed will be halved
13. A voltage $\mathrm{v}(\mathrm{t})=173 \sin \left(314 \mathrm{t}+10^{\circ}\right)$ is applied to a circuit. It causes a current flow described by

$$
i(t)=14.14 \sin \left(314 t-20^{\circ}\right)
$$

The average power delivered is nearly
[EC ESE - 2017]
(a) 2500 W
(b) 2167 W
(c) 1500 W
(d) 1060 W
7. Consider the following statements respect to a parallel R-L-C circuit:
1.The bandwidth of the circuit
2.The bandwidth of the circuit remain same if $L$ is increased.
3.At resonance, input impedance is a real quantity.
4.At resonance, the magnitude of the input impedance attains its $m$ inimum value.
Which of the above statements are correct?
[EC ESE - 2017]
(a) 1,2 and 4
(b) 1, 3 and 4

## CHAPTER - 4

MAGNETICALLY COUPLED CIRCUIT

### 4.1 INTRODUCTION

When two circuits are so placed that a portion of the magnetic flux produced by one links with the turns of both, they are said to be mutually coupled magnetically. This effect is characterized by mutually inductance (M)
Mutual Inductance (M) is the property of magnetic coupling showing an induction of voltage in one coil/winding by a change of current in other coil/winding.


In the above figure two coils 1 and 2 with turns $\mathrm{N}_{1}$ and $\mathrm{N}_{2}$ are placed close to each other so that part of flux of one coil links with other coil too. The current $i_{1}$ in coil 1 produces flux $\phi_{1}$. Some part of $\phi_{1}$ links only with coil 1 let this is $\phi_{11}$ this is known as self flux or leakage flux of coil 1. $\phi_{12}$ is the flux which links with both the coils. $\phi_{12}$ is called mutual flux. Similarly current $i_{2}$ in coil 2 produces $\phi_{2}$ which has $\phi_{22}$ and $\phi_{21}$ as its components. $\phi_{22}$ links only with coil 2 and $\phi_{21}$ links with both coils.
Now the voltage induced in coil 2 by change in current of coil $1 i_{1}$
$\mathrm{v}_{21}=\mathrm{M}_{21} \frac{\mathrm{di}_{1}}{\mathrm{dt}}$
However by Faraday's Law
$\mathrm{v}_{21}=\mathrm{N}_{2} \frac{\mathrm{~d} \phi_{12}}{\mathrm{dt}}$
$\Rightarrow \mathrm{M}_{21}=\mathrm{N}_{2} \frac{\mathrm{~d} \phi_{12}}{\mathrm{di}_{1}}$
$\Rightarrow \mathrm{M}_{21}=\mathrm{N}_{2} \frac{\mathrm{~d} \phi_{12}}{\mathrm{di}_{1}}$
If air is the medium between two coils, then magnetization is linear and
$\frac{\mathrm{d} \phi_{12}}{\mathrm{di}_{1}}=\frac{\phi_{12}}{\mathrm{i}_{1}}$
Hence $M_{21}=\frac{N_{2} \phi_{12}}{i_{1}}$
Similarly $M_{12}=\frac{N_{1} \phi_{21}}{i_{2}}$
Since the reluctance of both the fluxes i.e. $\phi_{12} \& \phi_{21}$ is same $M_{12} \& M_{21}$ are equal say $M_{12}=M_{21}=$ M.

## WORKBOOK

Example 1. The number of turns in two coupled coils are 600 and 1200 respectively. When a current of 4 A flows in coil 1 , the total flux in this coil is 0.5 m Wb and the flux linking coil 2 is 0.4 m Wb . Determine $\mathrm{L}_{1}, \mathrm{~L}_{2}, \mathrm{M}$ and k .

## Solution.

$\mathrm{L}_{1}=\frac{\phi_{1} \mathrm{~N}_{1}}{\mathrm{I}_{1}}=\frac{0.5 \times 10^{-3} \times 600}{4}=0.075 \mathrm{H}$
Since the self inductance is direction proportional to the square of the number of turns $\mathrm{L} \propto \mathrm{N}^{2}$ and
$\frac{\mathrm{L}_{1}}{\mathrm{~L}_{2}}=\frac{\mathrm{N}_{1}^{2}}{\mathrm{~N}_{2}^{2}}$
$\mathrm{L}_{2}=\left(\frac{\mathrm{N}_{2}}{\mathrm{~N}_{1}}\right)^{2} \times \mathrm{L}_{1}$
$=\left(\frac{1200}{600}\right)^{2} \times 0.075=0.3 \mathrm{H}$
$\mathrm{M}=\frac{\mathrm{N}_{2} \phi_{12}}{\mathrm{~L}_{1}}=\frac{1200 \times 0.4 \times 10^{-3}}{4}=0.12 \mathrm{H}$
$\mathrm{k}=\frac{\phi_{12}}{\phi_{1}}=\frac{0.4}{0.5}=0.8$
Alternatively, $\mathrm{k}=\frac{\mathrm{M}}{\sqrt{\mathrm{L}_{1} \mathrm{~L}_{2}}}=\frac{0.12}{\sqrt{0.075 \times 0.3}}=0.8$

Example 2. With reference to fig. Find $v_{x-y}$, if $\frac{\mathrm{di}}{\mathrm{dt}}=300 \mathrm{~A} / \mathrm{sec}$. Assume $\mathrm{L}_{1}=\mathrm{L}_{2}=1 \mathrm{H}$ and $M_{12}=\frac{1}{2} \mathrm{H}$.


## Solution.

Applying KVL in the respective loops,
$\mathrm{v}_{\mathrm{x}-\mathrm{n}}=\mathrm{L}_{1} \frac{\mathrm{di}}{\mathrm{dt}}-\mathrm{M}_{12} \frac{\mathrm{di}}{\mathrm{dt}}$
[As the coils are electrically in series, hence, the same current I passes through the coils; $M$ is negative]
Or $v_{x-n}=\frac{\mathrm{di}}{\mathrm{dt}}-0.5 \frac{\mathrm{di}}{\mathrm{dt}}=0.5 \frac{\mathrm{di}}{\mathrm{dt}}$
Similarly, $v_{y-n}=L_{2} \frac{d i}{d t}-M_{12} \frac{d i}{d t}$
$=1 . \frac{\mathrm{di}}{\mathrm{dt}}-0.5 \frac{\mathrm{di}}{\mathrm{dt}}=0.5 \frac{\mathrm{~d}}{\mathrm{dt}}$
$\therefore \quad v_{x-y}=v_{x-n}+v_{y-n}$
$=0.5 \frac{\mathrm{di}}{\mathrm{dt}}+0.5 \frac{\mathrm{di}}{\mathrm{dt}}=\frac{\mathrm{di}}{\mathrm{dt}}$
i.e., $\left|v_{x-y}\right|=300 \mathrm{~V}$

Example 3. Two impedances $\mathrm{Z}_{1}$ and $\mathrm{Z}_{2}$ are connected in series with the primary and secondary winding of an ideal transformer where the primary coil has $\mathrm{J} 2 \Omega$ and the secondary coil has $j 3 \Omega$ reactance. Find the mutual reactance and inductance if $\omega=100$ $\mathrm{rad} / \mathrm{sec}$.


Solution.
In ideal transformer, $\mathrm{K}=1$
$\therefore \mathrm{M}=\sqrt{\mathrm{L}_{1} \mathrm{~L}_{2}}$
i..e, $X_{M}=\sqrt{X_{1} X_{2}}$
$X_{1}$ is reactance of primary coil
$X_{2}$ is reactance of secondary coil
$X_{M}$ is mutual reactance
i.e. $X_{M}=\sqrt{j 2 \times j 3}=j 2.45 \Omega$
but $\mathrm{X}=2 \pi \mathrm{fL}$

## CHAPTER - 5

NETWORK THEOREMS

### 5.1 THEVENIN'S THEORM

Any two terminal bilateral linear circuit can be replaced by an equivalent circuit consisting of a voltage source and a series resistor.

### 5.1.1 Steps for Solving a Network using Thevenin's Theorem

1. Remove the load resistor $\left(\mathrm{R}_{\mathrm{L}}\right)$ and find the open circuit voltage $\left(\mathrm{V}_{\mathrm{oc}}\right)$ across the open circuited load terminals.
2. Find the Thevenin's resistance $\left(\mathrm{R}_{\mathrm{TH}}\right)$

## 1. If Circuit contains only Independent Sources

Deactivate the constant sources (for voltage source remove it by short circuit and for the current source remove it by open circuit) and find the internal resistance $\left(\mathrm{R}_{\mathrm{TH}}\right)$ of the source side looking through the open circuited load terminals.

## 2. For the circuits containing dependent sources in addition to or in absence of independent sources

Find $\mathrm{V}_{\mathrm{OC}}$ by open circuiting the load terminals. Then short the load terminals and find the short circuit current ( $\mathrm{I}_{\mathrm{SC}}$ ) through the shorted terminals.
Thevenin's resistance is given as: $\mathrm{R}_{\mathrm{TH}}=\frac{\mathrm{V}_{\mathrm{OC}}}{\mathrm{I}_{\mathrm{SC}}}$

(i) Obtain thevenin's equivalent circuit by placing $\mathrm{R}_{\mathrm{TH}}$ in series with $\mathrm{V}_{\mathrm{OC}}$
(ii) Reconnect $R_{L}$ across the load terminals.

### 5.1.2 Thevenin's Equivalent Network

$I($ Load current $)=\frac{V_{O C}}{R_{T H}+R_{L}}$


If only dependent sources are present in circuit, $\mathrm{R}_{\text {Th }}=\frac{\mathrm{V}_{\text {test }}}{1_{\text {test }}} ; \mathrm{I}_{\text {test }}=1 \mathrm{~A}$
$\mathrm{V}_{\text {test }}$ is calculated across the load by short circuiting it, and current of 1 A flows through the short circuited branch as $I_{\text {test }}$. Then $R_{T H}=V_{\text {test }}$

Dividing equation (ii) by equation (i)

$$
\begin{aligned}
& 3=\sqrt{\frac{624+\mathrm{C}_{\mathrm{s}}}{60.4+\mathrm{C}_{\mathrm{s}}}} \\
& \Rightarrow \quad 543.6+9 \mathrm{C}_{\mathrm{s}}=624+\mathrm{C}_{\mathrm{s}} \\
& \Rightarrow \quad 8 \mathrm{C}_{\mathrm{s}}=80.4 \\
& \Rightarrow \quad \mathrm{C}_{\mathrm{s}}=10.05 \mathrm{pF}
\end{aligned}
$$

Sol. 9. (a)
In an L-C function,
(i) Poles and zeros are alternate on $\mathrm{j} \omega$ axis.
(ii) There is either a pole or a zero at origin and infinity.
(iii) The highest and lowest powers of $s$ in numerator and denomenator can differ at the most by 1 .

## CHAPTER - 7

### 7.1 LAPLACE TRANSFORMATION

The Laplace transformation of a function $f(\mathrm{t})$ is defined as
$\mathrm{F}(\mathrm{s})=\operatorname{Lf}(\mathrm{t})=\int_{0}^{\infty} \mathrm{f}(\mathrm{t}) \mathrm{e}^{-\mathrm{st}} \mathrm{dt}$
Where as in complex frequency
$s$ being the intermediate or transformation variable.
7.1.1 Laplace Transform of a Derivative $\left[\frac{d f(t)}{d t}\right]$
$L f^{\prime}(t)=s F(s)-f(0+)$
7.1.2 Laplace Transform of an Integral $\int f(t) d t$
$L\left(\int f(t) d t\right)=\left.\frac{1}{s} \int f(t) d t\right|_{0+}+\frac{1}{s} F(s)$
$\left[\left.\int f(t)\right|_{0+}\right.$ gives the value of the integral at $\left.t=0+\right]$

### 7.1.3 Frequency Shifting

$L\left(e^{a t} f(t)\right)=F(s-a)$
$L\left(e^{-a t} f(t)\right)=F(s+a)$

### 7.2 LAPLACE TRANSFORM OF COMMON FORCING FUNCTIONS

| $\boldsymbol{f ( t )}$ | $\mathbf{F}(\mathbf{s})$ | $\boldsymbol{f ( t )}$ | $\mathbf{F ( s )}$ |
| :--- | :--- | :--- | :--- |
| $u(t)$ | $\frac{1}{\mathrm{~s}}$ | $\mathrm{e}^{-\alpha t^{n}}$ | $\frac{\underline{\mathrm{n}}}{(\mathrm{s}+\alpha)^{\mathrm{n}+1}}$ |
| $\mathrm{e}^{-\alpha \mathrm{t}}$ | $\frac{1}{\mathrm{~s}+\alpha}$ | $\mathrm{e}^{-\alpha \mathrm{t}} \sin \omega \mathrm{t}$ | $\frac{\omega}{(\mathrm{s}+\alpha)^{2}+\omega^{2}}$ |
| $\sin \omega \mathrm{t}$ | $\frac{\omega}{\mathrm{s}^{2}+\omega^{2}}$ | $\mathrm{e}^{-\alpha \mathrm{t}} \cos \omega \mathrm{t}$ | $\frac{\mathrm{s}+\mathrm{a}}{(\mathrm{s}+\alpha)^{2}+\omega^{2}}$ |
| $\cos \omega \mathrm{t}$ | $\frac{\omega}{\mathrm{s}^{2}+\omega^{2}}$ | $\delta(\mathrm{t})$ | 1 |
| t | $\frac{1}{\mathrm{~s}^{2}}$ | $\operatorname{Sinh} \theta \mathrm{t}$ | $\frac{\theta}{\mathrm{s}^{2}-\theta^{2}}$ |

## CHAPTER - 8

RESONANCE

### 8.1 RESONANCE

Resonance in electrical circuits consisting of passive and active elements represents a particular state of the circuit when the current or voltage in the circuit is maximum or minimum with respect to the magnitude of excitation at a particular frequency, the circuit impedance being either minimum of maximum at the power factor unity.
The phenomenon of resonance is observed in both series or parallel a.c. circuits comprising of $R$, L and C and excited by an a.c. source.

### 8.2 SERIES RESONANCE


$Z_{\text {in }}=\frac{V}{I}=R+j\left(\omega L-\frac{1}{\omega c}\right)$
For resonance V \& I must be in same phase
So for some frequency $\omega=\omega_{0}$
$Z_{\text {in }}=R+j_{0} \Rightarrow \omega_{0} L-\frac{1}{\omega_{0} C}=0 \Rightarrow \omega_{0}=\frac{1}{\sqrt{\text { LC }}}$
$I=\frac{V}{|z|}=\frac{V}{\sqrt{R^{2}+(\omega \mathrm{L}-1 / \omega \mathrm{c})^{2}}}$ at $\omega_{0}, I=\frac{\mathrm{V}}{\mathrm{R}}$


Points A \& B are half power or 3 dB points because $20 \log _{10}\left(\frac{1}{2}\right)=3 \mathrm{~dB}$
Band width of circuit $\Delta \omega=\mathrm{BW}=\omega_{2}-\omega_{1}$
Quality factor
$Q_{0}=2 \pi\left[\frac{\text { Max energy stored }}{\text { Total energy last per perior }}\right]$
$\mathrm{Q}_{0}=2 \pi\left[\frac{\omega_{\mathrm{L}}+\omega_{\mathrm{C}}}{\mathrm{P}_{\mathrm{R}} \mathrm{T}}\right]$

## CHAPTER - 9

TWO PORT NETWORKS

### 9.1 INTRODUCTION

The terminal pair is called as a "port". If the current entering one terminal of a pair is equal and opposite to the current leaving the other terminal of the pair.


### 9.2 TWO-PORT NETWORK

A two-port network is shown, by which we observe that a two-port network is represented by a black box with four variables, namely, two voltages $\left(V_{1}, V_{2}\right)$ and two currents $\left(I_{1}, I_{2}\right)$ which are available for measurements and are relevant for the analysis of two port networks. Of these four variables which two variable may be considered `independent` and which two `dependent` is generally decided by the probable under consideration

| Two Port Parameters |  |  |  |
| :---: | :---: | :---: | :---: |
| Name | Express | In terms of | Matrix Equation |
| Open circuit impedance [Z] | $\mathrm{V}_{1}, \mathrm{~V}_{2}$ | $\mathrm{I}_{1}, \mathrm{I}_{2}$ | $\left[\begin{array}{l}\mathrm{V}_{1} \\ \mathrm{~V}_{2}\end{array}\right]=\left[\begin{array}{ll}\mathrm{Z}_{11} & \mathrm{Z}_{12} \\ \mathrm{Z}_{21} & \mathrm{Z}_{22}\end{array}\right]\left[\begin{array}{l}\mathrm{I}_{1} \\ \mathrm{I}_{2}\end{array}\right]$ |
| Short-circuit admittance [Y] | $\mathrm{I}_{1}, \mathrm{I}_{2}$ | $\mathrm{V}_{1}, \mathrm{~V}_{2}$ | $\left[\begin{array}{l}\mathrm{I}_{1} \\ \mathrm{I}_{2}\end{array}\right]=\left[\begin{array}{lll}\mathrm{Y}_{11} & \mathrm{Y}_{12} \\ \mathrm{Y}_{21} & \mathrm{Y}_{22}\end{array}\right]\left[\begin{array}{l}\mathrm{V}_{1} \\ \mathrm{~V}_{2}\end{array}\right]$ |
| Transmission or  <br> Chain $\quad[\mathrm{T}]$ or  <br> $[\mathrm{ABCD}]$   | $V_{1}, I_{2}$ | $\mathrm{V}_{2}, \mathrm{I}_{2}$ | $\left[\begin{array}{l}\mathrm{V}_{1} \\ \mathrm{I}_{1}\end{array}\right]=\left[\begin{array}{ll}\mathrm{A} & \mathrm{B} \\ \mathrm{C} & \mathrm{D}\end{array}\right]\left[\begin{array}{c}\mathrm{V}_{2} \\ -\mathrm{I}_{2}\end{array}\right]$ |
| Inverse Transmission [ $\mathrm{T}^{\prime}$ ] | $\overline{\mathrm{V}_{2}, \mathrm{I}_{2}}$ | $\mathrm{V}_{1},-\mathrm{I}_{1}$ | $\left[\begin{array}{l}\mathrm{V}_{2} \\ \mathrm{I}_{2}\end{array}\right]=\left[\begin{array}{l}\mathrm{A}^{\prime} \mathrm{B}^{\prime} \\ \mathrm{C}^{\prime} \mathrm{D}^{\prime}\end{array}\right]\left[\begin{array}{r}\mathrm{V}_{1} \\ -\mathrm{I}_{1}\end{array}\right]$ |
| Hybrid (h) | $\mathrm{V}_{1}, \mathrm{I}_{2}$ | $\mathrm{I}_{1}, \mathrm{~V}_{2}$ | $\left[\begin{array}{l}\mathrm{V}_{1} \\ \mathrm{I}_{2}\end{array}\right]=\left[\begin{array}{ll}\mathrm{h}_{11} & \mathrm{~h}_{12} \\ \mathrm{~h}_{21} & \mathrm{~h}_{22}\end{array}\right]\left[\begin{array}{l}\mathrm{I}_{1} \\ \mathrm{~V}_{2}\end{array}\right]$ |
| Inverse hybrid (g) | $\mathrm{I}_{1}, \mathrm{~V}_{2}$ | $\mathrm{V}_{1}, \mathrm{I}_{2}$ | $\left[\begin{array}{l}\mathrm{I}_{1} \\ \mathrm{~V}_{2}\end{array}\right]=\left[\begin{array}{ll}\mathrm{g}_{11} & \mathrm{~h}_{12} \\ \mathrm{~g}_{21} & \mathrm{~h}_{22}\end{array}\right]\left[\begin{array}{l}\mathrm{V}_{1} \\ \mathrm{I}_{2}\end{array}\right]$ |

### 9.3 OPEN CIRCUIT IMPEDANCE (Z) PARAMETERS

Expressing two-port voltages in terms of two-port currents
$\left(\mathrm{V}_{1}, \mathrm{~V}_{2}\right)=\mathrm{f}\left(\mathrm{I}_{1}, \mathrm{I}_{2}\right)$
$\left[\begin{array}{l}\mathrm{V}_{1} \\ \mathrm{~V}_{2}\end{array}\right]=\left[\begin{array}{ll}\mathrm{Z}_{11} & \mathrm{Z}_{12} \\ \mathrm{Z}_{21} & \mathrm{Z}_{22}\end{array}\right]\left[\begin{array}{l}\mathrm{I}_{1} \\ \mathrm{I}_{2}\end{array}\right]$ or $[\mathrm{V}]=[\mathrm{Z}][\mathrm{I}]$

### 10.1 IMPORTANT DEFINITIONS

1. Graph: It is the collection of nodes and Branch of a network.

(A)

(B)
2. Branch: Each oriented line segment of the graph is called branch.
3. Node: The end point of a branch is called node.
4. Incident Branch: Branch whose end fall on a node is called incident branch.

## 5. Connected and Non-Connected Graph

If there exists a path between every pair of nodes of a graph, then the graph is called connected graph, otherwise graph is called non-connected graph.
6. Degree of Node: Degree of Node is the number of branches incident on the node.
7. Subgraph: A portion of graph is called subgraph
8. Path: Path is transverse from one node to another node
9. Loop: Loop is a collection of branches in a graph which form a closed path.
10. Tree: The collection of minimum no. of branches connecting all the nodes of a graph without making a loop.
A single graph can have many no. of trees.
The no. of trees for a given graph $=\mathrm{n}-1$
where $\mathrm{n} \rightarrow$ no. of nodes
11. Twig: Branch of a tree is called a twig.
12. Cotree: Remaining part of a graph after removal of twigs is called cotree. It is collection of links.
13. Links: are the branches removed from the graph to make a tree.

Total no. of branch of a graph are given by $b=(n-1)+L$
n is no. of nodes
L is No. of links
No. of twigs $=(n-1)=$ no. of KCL equation

## CHAPTER - 11

## NETWORK FUNCTIONS

### 11.1 INTRODUCTION

The basic definition of one port and two port network being discussed earlier, here we will discuss about the transform of excitation and response along with their relations. A network function exhibits the relationship between the transform of the source or excitation to the transform of the response for a electrical network. Further to this, we will discuss the stability of the network function mathematically formulating the network function mathematically formulating the network function through "pole-zero" concept.
11.2 DRIVING POINT IMPEDANCE AND ADMITTANCE

The driving point impedance of a one port network is defined as
$\mathrm{Z}(\mathrm{s})=\frac{\mathrm{V}(\mathrm{s})}{\mathrm{I}(\mathrm{s})}$
While the driving point admittance is given as
$\mathrm{Y}(\mathrm{s})=\frac{\mathrm{I}(\mathrm{s})}{\mathrm{V}(\mathrm{s})}$
For the one port network
Similarly, for the two port network, the driving point impedance and admittance at port 1 is defined as
$\mathrm{Z}_{11}(\mathrm{~s})=\frac{\mathrm{V}_{1}(\mathrm{~s})}{\mathrm{I}_{1}(\mathrm{~s})}$
and $\mathrm{Y}_{11}(\mathrm{~s})=\frac{\mathrm{I}_{1}(\mathrm{~s})}{\mathrm{V}_{1}(\mathrm{~s})}$
While the driving point impedance and admittance at the port 2 are designated as
$\mathrm{Z}_{22}(\mathrm{~s})=\frac{\mathrm{V}_{2}(\mathrm{~s})}{\mathrm{I}_{2}(\mathrm{~s})}$
and $Y_{22}(\mathrm{~s})=\frac{\mathrm{I}_{2}(\mathrm{~s})}{\mathrm{V}_{2}(\mathrm{~s})}$

### 11.3 TRANSFER IMPEDANCE AND ADMITTANCE

Transfer impedance is defined as the ratio of transform voltage at output port to the transformed current at the input port of a two port network.
This gives, $Z_{12}(s)=\frac{V_{2}(s)}{I_{1}(s)}$
In a similar way, the transfer admittance is defined as the ratio of current transform at output port to the voltage transform at the input port. It is given as
$\mathrm{Y}_{12}(\mathrm{~s})=\frac{\mathrm{I}_{2}(\mathrm{~s})}{\mathrm{V}_{1}(\mathrm{~s})}$

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