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GATE 2019

NETWORK ANALYSIS

ELECTRONICS ENGINEERING





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GATE-2019: Network Analysis | Detailed theory with GATE & ESE previous year papers and detailed solu ons.

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CHAPTER - 1 BASIC CONCEPTS

1.1INTRODUCTION

1.1.1 Charge

Charge can be classified as:

- 1. Stationary Charge
- 2. Dynamic Charge

1. Stationary Charge

Stationary charge does not result into electric current because the flow of current means charge moving with net rate across any cross section.

(i) Any electric circuit should always follow law of conservation of charge and law of conservation of energy.

(ii) Circuit theory is analysed always at low frequency and field theory always at high frequency.(iii) Transit time effect is always neglected at low frequency because

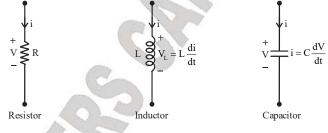
 $T >> t_r$

Where T is time period of sinosdual signal

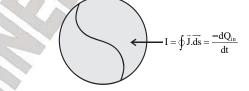
tr is Transit Time (time taken by signal effect to travel from one point to another point).

(iv) Elemental law is obeyed only at low frequency such as ohm's law. It is not applicable at high frequency because of distributed nature of element.

(v) Elemental law always depend upon the nature of element



For different Element, Different Form of Ohm's Law is present (i) In time domain, the ohm's law are applicable and also in frequency domain.



Current flowing out of this body is given by equation of continuity as below

$$I = \oint \vec{J} \cdot \vec{ds} = -\frac{dQ_{in}}{dt} \qquad \dots (i)$$

This equation gives the law of conservation of charge.

If $\frac{dQ_{in}}{dt} = 0$; means no rate of charge of charge within body then eq.(i) become

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Example 1. The waveform of the current through an inductor of 10H is shown in Fig. Sketch the waveform of the voltage across the inductor

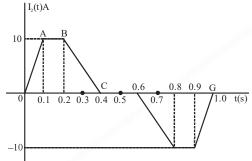


Fig. (a) Waveform of i

Solution.

The voltage across an inductor is

$$v_L = L \frac{di_L}{dt}$$

(a) Between 0 and 0.1 s (that is, for 0 < t < 0.1), the derivative, $\frac{di_L}{dt}$, which is the slope of the i_L

curve, is constant since $i_{I}(t)$ is linear

$$\frac{di_{L}}{dt} = \text{slope of line OA}$$
$$= \frac{OA}{AM} = \frac{10}{0.1} = 100 \text{ A/s (constant)}$$

AM 0.1

$$\therefore$$
 v_L = L $\frac{di_L}{dt}$ = 10×100 = 1000V (constant) for

0 < t < 0.1This constant value is plotted as a horizontal line ab in fig.

(b) For 0.1 < t < 0.2s the current curve AB is horizontal, that is, its slope is zero

$$\therefore v_{L} = L\frac{di}{dt} = 10 \times 0 = 0V$$

For 0.1 < t < 0.2s
This is plotted as horizontal li

ine cd in fig. (c) For 0.2 < t < 0.4s, the current curve is BC. The slope of BC is negative.

$$\therefore \frac{di_{L}}{dt} = \text{slope of line BC}$$

$$= \frac{BN}{NC} = \frac{10}{0.2} = -50 \text{ A/s}$$

$$\therefore v_{L} = L \frac{di_{L}}{dt} = 10 \times 0 = 0 \text{ for } 0.4 < t < 0.6s$$
The voltage waveform in the range $0.4 < t < 0.6s$
The voltage waveform in the range $0.4 < t < 0.6s$ is by fh in Fig. (b)
e) For $0.6 < t < 0.8s$, the current curve is DE. I has a negative slope given by
$$\frac{di_{L}}{dt} = -\frac{EP}{PD} = -\frac{10}{0.2} = -50 \text{ A/s}$$

$$\therefore v_{L} = L \frac{di_{L}}{dt} = 10 \times (-50) = -500 \text{ V} \text{ for } 0.6 < t < 0.8s$$
The voltage waveform in this range is show by cm in fig.
f) For $0.8 < t < 0.9s$, since the current curve EF

horizontal,
$$\frac{di_L}{dt} = 0$$

 $v_L = L \frac{di_L}{dt} = 10 \times 0 = 0V$

is

This is shown by curve ℓ n in fig. 9b)

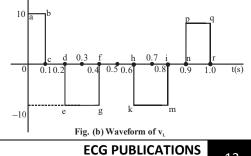
(g) for 0.9 < t < 1.0s, the current curve FG has a slope

$$\frac{di_{L}}{dt} = \frac{FQ}{QG} = \frac{10}{0.1} = 100 \text{A/s}$$

:.
$$v_{\rm L} = L \frac{{\rm d} {\rm I}_{\rm L}}{{\rm d} t} = 10 \times 100 = 1000 {\rm V}$$

This is shown by curve pq in Fig.(b) Overall voltage waveform is shown below:

 $I_2(t)A$

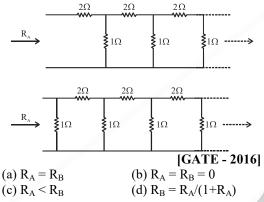




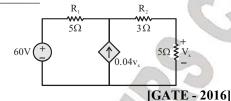


GATE QUESTIONS

as shown below. The circuits extend infinitely (c) Current controlled current source in the direction shown. Which one of the (d) Current controlled voltage source statements is TRUE?



2. In the circuit shown in the figure, the magnitude of the current (in amperes) through R₂ is



3. An incandescent lamp is marked 40W, 240V. If resistance at room temperature (26°C) is 120 Ω , and temperature coefficient of resistance is $4.8 \times 10^{-3/\circ}$ C, then its 'ON' state filament temperature in °C is approximately

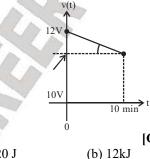
[GATE - 2014]

4. The circuit shown in the figure represents a 4 8 2 3 5 6 7 t(µs) ≷R A₁I∢ will be [GATE - 2014] (a) 8nC (b) 10 nC (a) Voltage controlled voltage source (c) 13 nC (d) 16 nC

1. R_A and R_B are the input resistances of circuits (b) Voltage controlled current source

G

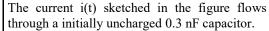
5. A fully charged mobile phone with a 12V battery is good for a 10 minute talk-time. Assume that during the talk - time the battery delivers a constant current of 2A and its voltage drops linearly from 12V to 10V as shown in the figure. How much energy does the battery deliver during the talk - time?

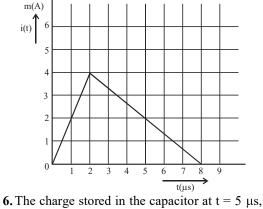


[GATE - 2009]

(a) 220 J (c) 13.2 kJ (d) 14.4 J

Common data for Q. 6 & Q. 7

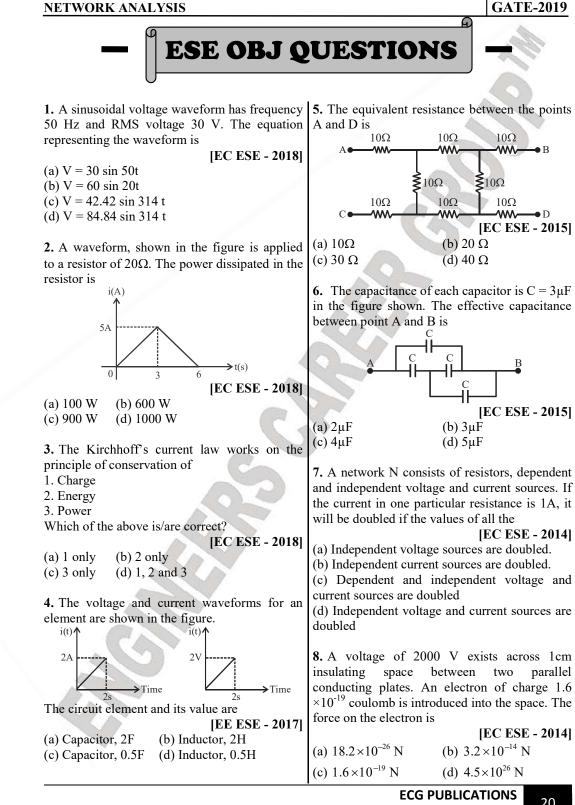




[GATE - 2008]

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CHAPTER - 2 NETWORK LAWS

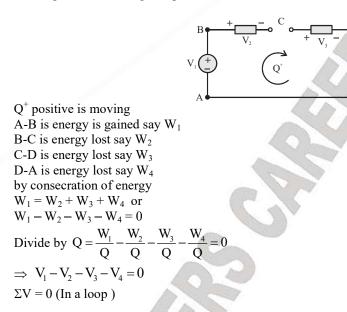
2.1 KIRCHOFF'S VOLTAGE LAW (KVL)

It states that algebraic sum of all voltages in a closed path or loop is zero

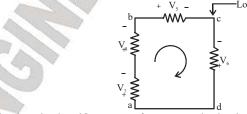
$$\sum_{\text{loop}} V = 0$$

For writing KVL start from any point in the loop and come to the same point via transversing the path of closed loop. While doing so take voltage rises as positive and voltage drops as negative then

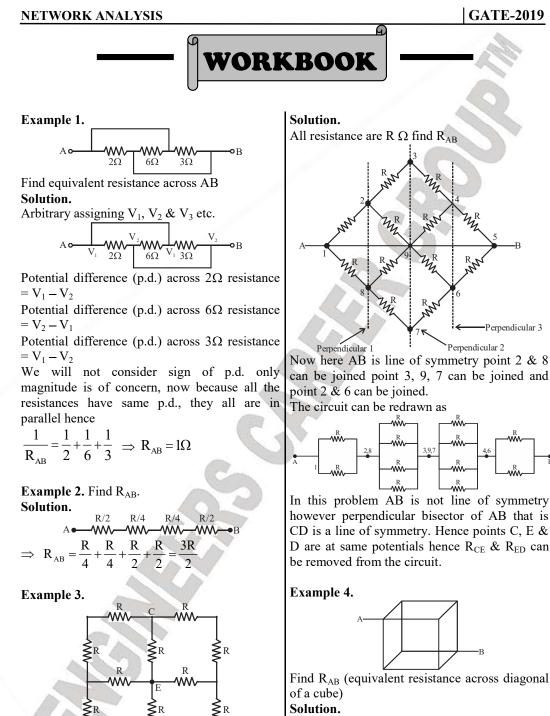
 Σ voltage rise + Σ voltage drops = 0



Example. KVL in this loop starting from a in clockwise direction is $-V_7 - V_8 - V_5 + V_6 = 0$ $\Rightarrow V_6 = V_7 + V_8 + V_5$



The basis of the law is that if we start from a particular junction and go round the mesh till we come back to the starting point, then we must be at the same potential with which we started. Hence it means that all the sources of e.m.f. met on the way must necessarily be equal to the voltage drops in the resistances, every voltage being given its proper sign, plus or minus.

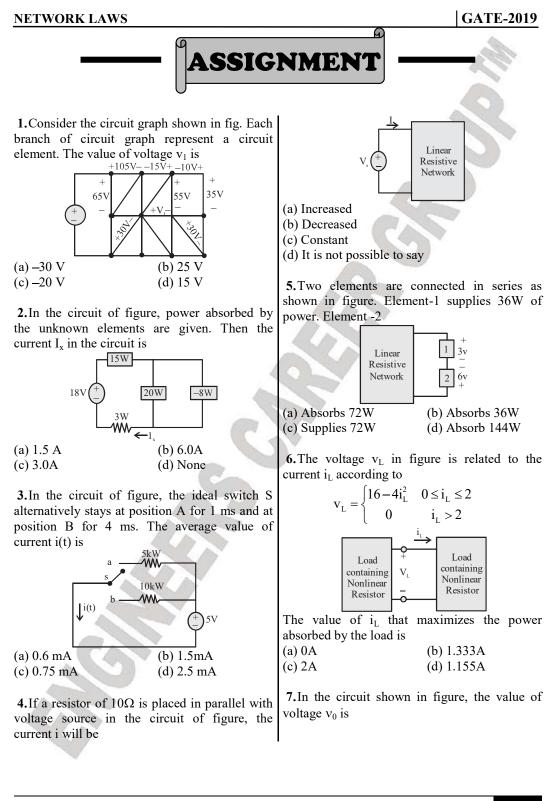


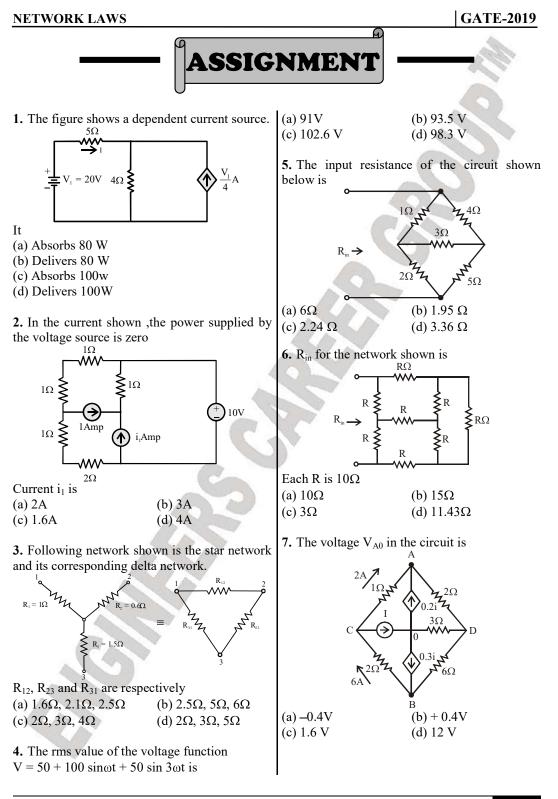
R

Find equivalent resistance between AB

For solving such problems. Assume current I is entering at A then this i will come out of B then this current is distributed as per symmetry

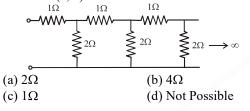
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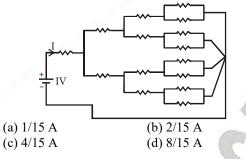




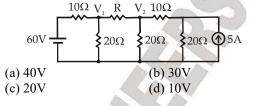
1. The equivalent resistance seen between the (b)Voltage source of 25V with +ve terminal terminals (a, b) is: downward



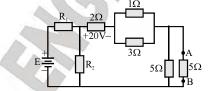
2. All the resistances in the figure are 1 Ω each. The value of I will be:



3. In the circuit below, $V_1 = 40$ V when R is 10 Ω , when R is zero, the value of V₂ will be:



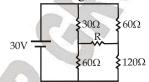
4. In the circuit shown below, the voltage across 2Ω resistor is 20V. The 5Ω resistor connected between the terminals A and B can be replaced by an ideal.



(a)Voltage source of 25V with +ve terminal upward

(c)Current source of 2A upward. (d)Current source of 2A downward.

5. Consider the following circuit:



What is the power delivered to resistor in the above circuit?

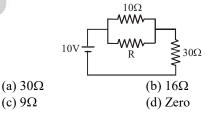
$$(a) - 15 W$$

(b) 0W

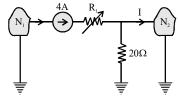
(c) 15 W

(d) Cannot be determined unless the value of R is known.

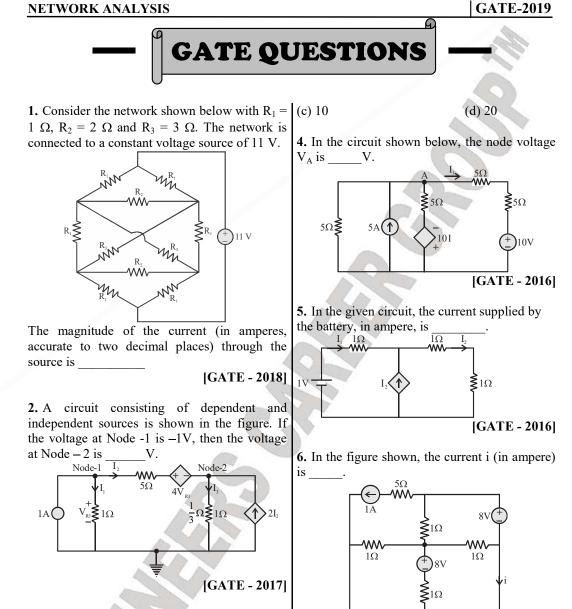
6. In the circuit shown in the figure, the power dissipated in 30Ω resistor will be maximum if the value of R is:



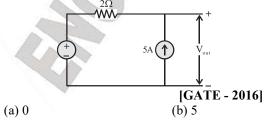
7. In the circuit shown in the figure, for R = 20Ω , the circuit 'I' is 2A. When R is 10Ω , the current 'I' would be:



8. In the figure below, the current of 1A flows through the resistance of:



3. In the circuit shown below, the voltage and current sources are ideal. The voltage (V_{out}) across the current source, in volts, is



7. In the given circuit, each resistor has a value equal to 1Ω

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[GATE - 2016]

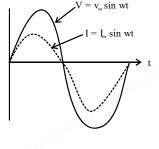
NETWORK LAWS

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ESE OBJ QUESTIONS 3. Lenz's law is direct consequence of the law 1. Statement I: Two ideal current sources with currents I_1 and I_2 of conservation of energy. cannot be connected in parallel. Which of the above statements are correct? **Statement II:** [EC ESE - 2017] Superposition theorem cannot be applied to (a) 1 and 2 only (a)(b) 1 and 3 only ideal current sources if these sources are (c) 2 and 3 only (d) 1, 2 and 3 connected in cascade. [EC ESE - 2017] 4. Consider the following factors: **Codes:** 1. Number of turns of the coil (a) Both Statement I and II are individually true 2. Length of the coil and Statement II is the correct explanation of 3. Rea of cross-section of the coil 4. permeability of the core Statement I. (b) Both Statement I and Statement II are On which of the above factors does inductance, individually true but Statement II is not the depend? correct explanation of Statement I. [EC ESE - 2017] (c) Statement I is true but Statement II is false. (a) 1, 2 and 3 only (b) 1, 3 and 4 only (d) Statement I is false but Statement II is true. (c) 1, 2, 3, and 4 (d) 2 and 4 only **2.** What is the current through the 8Ω resistance 5. Consider a packet switched network based on connected across terminals, M and N in the a virtual circuit mode of switching. The delay jutter for the packets of a session from the circuit? source node to the destination node is/are 12Ω 80 1. Always zero 2. Non-zero 3. For some networks, zero Select the correct answer using the code given 8ν 80 below. [EC ESE - 2017] (a) 1 (b) 2 only (c) 3 only (d) 2 and 3 6. A network in which all the elements are N physically separable is called a [EE ESE - 2017] (a) 0.34 A from M to N [EC ESE - 2017] (a) Distributed network (b) 0.29 A from M to N (b) Lumped network (c) 0.29 A from n to M (d) 0.34 A from N to M (c) Passive network (d) Reactive network **3.** Consider the following statements: 1. Flaming's rule is used where induced e.m.f is 7. For the active network shown in figure, the due to flux cutting. value of V/I is 2. Leng'z law is used when the induced e.m.f is due to change in flux linkages.

CHAPTER - 3 A.C ANALYSIS

3.1 AC THROUGH PURE OHMIC RESISTANCE ALONE



 $v = V_m \sin \omega t$ v = iR

$$i = \frac{V_m}{R} \sin \omega t$$

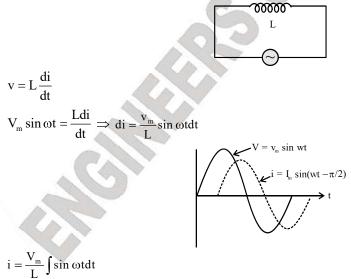
Current is max when $\sin wt = 1$

i.e.
$$I_m = \frac{V_m}{R}$$

 $\therefore i = I_m \sin \omega t$

3.2 AC THROUGH PURE INDUCTANCE ALONE

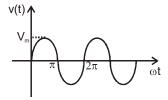
Whenever an alternating voltage is applied to a purely inductive coil, a back emf is produced due to the self-inductance of the coil





=

sine wave



 $v(t) = v_m \sin \omega t$

$$v_{av} = \frac{v_m}{2\pi} \int_0^{2\pi} v_m \sin \omega t \ d(\omega t) = 0$$

Average value of full cycle of sine and cosine is zero.

Average value for half cycle

$$v_{av} = \frac{V_m}{\pi} \int_0^{\pi} v_m \sin \omega t \ d(\omega t)$$

= $\frac{V_m}{\pi} \left[-\cos \omega t \right]_0^{\pi} = \frac{V_m}{\pi} \left[-\cos \pi + \cos 0 \right] = \frac{2V_m}{\pi}$
RMS value
 $V_{rms} = \sqrt{\frac{1}{\pi}} \int_0^{\pi} V_m^2 \sin^2 \omega t \ d(\omega t)$
= $V_m \sqrt{\frac{1}{\pi}} \int_0^{\pi} \frac{1}{2} (1 - \cos 2\omega t) \ d(\omega t)$
= $\frac{V_m}{\sqrt{2}}$

Example 2. Find average and RMS value of half wave rectified sine wave Solution.

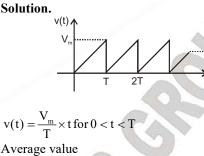




Average value $V_{m} = \frac{1}{\sqrt{\pi}} \int_{-\pi}^{\pi} V_{m} \sin \omega t \, d(\omega t) = \frac{V_{m}}{\sqrt{\pi}}$

$$V_{\text{rms}} = \sqrt{\frac{1}{2\pi} \int_0^{\pi} V_{\text{m}}^2 \sin^2 \omega t \, d(\omega t)} = \frac{V_{\text{m}}}{2}$$

Example 1. Average and RMS value of full Example 3. Find out average and RMS value of saw tooth wave form



$$V_{av} = \frac{1}{T} \int_{0}^{T} \frac{V_{m}}{T} t dt$$

$$= \frac{V_{m}}{T^{2}} \frac{t^{2}}{2} \bigg|_{0}^{T} = \frac{V_{m}}{2}$$

RMS value $V_{rms} = \sqrt{\frac{1}{T} \int_{0}^{T} \frac{V_{m}^{2}}{T^{2}} t^{2} dt}$
$$= \sqrt{\frac{V_{m}^{2}}{T^{3}} \int_{0}^{T} t^{2} dt}$$

 $\frac{V_m^2}{T^3} \frac{t^3}{3} \Big|_0^T = \frac{V_m}{\sqrt{3}}$

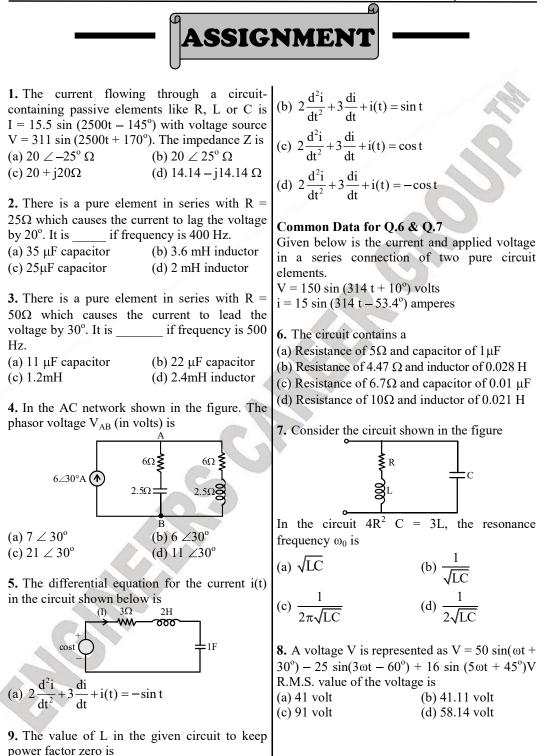
Example 4. Given $i_1(t) = 4\cos(\omega t + 30^\circ)$ A and $i_2(t) = 5\sin(\omega t - 20^\circ) \text{ A, find their sum.}$

Solution.

$$I_1 = 4 \angle 30^\circ$$

 $i_2 = 5 \cos (\omega t - 20^\circ - 90^\circ)$
 $= 5 \cos (\omega t - 110^\circ)$ And its phasor is
 $I_2 = 5 \angle -110^\circ$
 $I = I_1 + I_2$
 $= 4 \angle 30^\circ + 5 \angle -110^\circ$
 $= 3.464 + 2j - 1.71 - j4.698$
 $= 1.754 - j2.698$
 $= 3.218 \angle -56.97^\circ A$
Transforming this to time domain, we get

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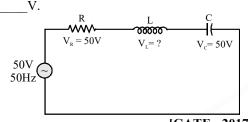


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GATE QUESTIONS

50Hz sinusoidal source. The voltages across the resistance and the capacitance are shown in the figure. The voltage across the inductor (V_L) is

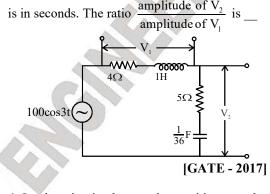


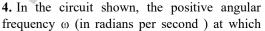
[GATE - 2017]

2. A connection is made consisting of resistance A is series with a parallel combination of resistance B and C. Three resistors of value 10 Ω , 5 Ω , 2 Ω are provided. Consider all possible permutations of the given resistors into the positions A, B, C and identify the configuration with maximum possible overall resistance: and also the ones with minimum possible overall resistance; and also the ones with minimum possible overall resistance. The ratio of maximum to minimum values of the resistances (up to second decimal place) is

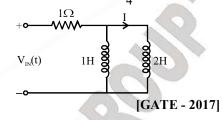
[GATE - 2017]

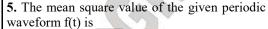
3. The figure shows an RLC circuit excited by the sinusoidal voltage 100cos (3t) Volts, where t

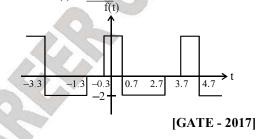




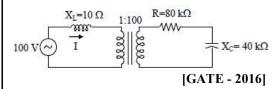
1. A series R-L-C circuit is excited with a 50v, the magnitude of the phase difference between the voltages V_1 and V_2 equals radians, is







6. The following figure shows the connection of an ideal transformer with primary to secondary turns ratio of 1:100. The applied primary voltage is 100V (rms), 50 Hz, AC. The rms value of the current I, in ampere, is



7. A resistance and a coil are connected in series and supplied from a single phase, 100 V, 50 Hz ac source as shown in the figure below. The rms values of plausible voltages across the resistance (V_R) and coil (V_C) respectively, in volts, are

ESE OBJ QUESTIONS 1. A two-element series circuit is connected **4**. One of the basic characteristics of any steady across an AC source given by state sinusoidal response of a linear R-L-C circuit with constant R, L and C values is $e = 200\sqrt{2}\sin(314t + 20)V$. The current is then [EC ESE - 2017] found to be $i = 10\sqrt{2}\cos(314t - 25)V$. The (a)The output remains sinusoidal with its parameters of the circuit are frequency being the same as that of the source [EE ESE - 2017] (b)The output remains sinusoidal with its (a) $R = 20 \Omega$ and $C = 160 \mu F$ frequency differing from that of the source (c)The output amplitude equals the soruce (b) $R = 14.14 \Omega$ and $C = 225 \mu F$ amplitude (c) L = 45 mH and $C = 225 \mu F$ (d)The phase angle difference between the (d) L = 45 mH and $C = 160 \mu F$ source and the output is always zero. **2.** Two resistors of 5Ω and 10Ω and an inductor 5. Three identical impedances are first L are connected in series across a 50 cos at connected in delta across a 3-phase balanced voltage source. If the power consumed by the supply. If the same impedances are now 5Ω resistor is 10W, the power factor of the connected in star across the same supply, then circuit is [EC ESE - 2017] [EE ESE - 2017] (a) The phase current will be one - third (b) 0.8 (a) 1.0 (b)The line current will be one - third (c) 0.6(d) 0.4 (c)The power consumed will be one -third (d)The power consumed will be halved 3. Statement (I): One series RC circuit and the other series RL circuit are connected in parallel 6. A voltage $v(t) = 173 \sin (314t + 10^{\circ})$ is across at ac supply. The circuit exhibits two applied to a circuit. It causes a current flow reasonance when L is variable. described by Statement (II): The circuit has two values of l $i(t) = 14.14 \sin (314t - 20^{\circ})$ for which the imaginary part of the input The average power delivered is nearly admittance of the circuit is zero. [EC ESE - 2017] [EE ESE - 2017] (a) 2500 W (b) 2167 W (a)Both Statement (I) and Statement (II) are (c) 1500 W (d) 1060 W individually true and Statement (II) is the correct explanation of Statement (I) 7. Consider the following statements respect to (b)Both Statement (I) and Statement (II) are a parallel R-L-C circuit: individually true but Statement (II) is not the 1.The bandwidth of the circuit correct explanation of Statement (I) 2. The bandwidth of the circuit remain same if L (c)Statement (I) is true but Statement (II) is is increased. false. 3.At resonance, input impedance is a real (d)Statement (I) is false but Statement (II) is quantity. true 4.At resonance, the magnitude of the input impedance attains its m inimum value. Which of the above statements are correct? [EC ESE - 2017] (a) 1, 2 and 4 (b) 1, 3 and 4

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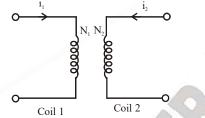
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СН MAGNETICALLY COUPLED C

4.1 INTRODUCTION

When two circuits are so placed that a portion of the magnetic flux produced by one links with the turns of both, they are said to be mutually coupled magnetically. This effect is characterized by mutually inductance (M)

Mutual Inductance (M) is the property of magnetic coupling showing an induction of voltage in one coil/winding by a change of current in other coil/winding.



In the above figure two coils 1 and 2 with turns N_1 and N_2 are placed close to each other so that part of flux of one coil links with other coil too. The current i_1 in coil 1 produces flux ϕ_1 . Some part of ϕ_1 links only with coil 1 let this is ϕ_{11} this is known as self flux or leakage flux of coil 1. ϕ_{12} is the flux which links with both the coils. ϕ_{12} is called mutual flux. Similarly current i_2 in coil 2 produces ϕ_2 which has ϕ_{22} and ϕ_{21} as its components. ϕ_{22} links only with coil 2 and ϕ_{21} links with both coils.

Now the voltage induced in coil 2 by change in current of coil 1 i_1

$$v_{21} = M_{21} \frac{dt_1}{dt}$$

However by Faraday's Law
$$v_{21} = N_2 \frac{d\phi_{12}}{dt}$$
$$\Rightarrow M_{21} = N_2 \frac{d\phi_{12}}{dt_1}$$
$$\Rightarrow M_{21} = N_2 \frac{d\phi_{12}}{dt_1}$$

If air is the medium between two coils, then magnetization is linear and dφ.

$$\frac{d\psi_{12}}{di_1} =$$

Hence M₂₁

di

Similarly
$$M_{12} = \frac{N_1 \phi_{21}}{i_2}$$

Since the reluctance of both the fluxes i.e. ϕ_{12} & ϕ_{21} is same M_{12} & M_{21} are equal say $M_{12} = M_{21} =$ М.



Example 1. The number of turns in two coupled coils are 600 and 1200 respectively. When a current of 4A flows in coil 1, the total flux in this coil is 0.5m Wb and the flux linking coil 2 is 0.4m Wb. Determine L_1, L_2, M and k. **Solution.** Applying $v_{x-n} = L_1$ [As the co

Solution.

$$L_1 = \frac{\phi_1 N_1}{I_1} = \frac{0.5 \times 10^{-3} \times 600}{4} = 0.075 H$$

Since the self inductance is direction proportional to the square of the number of turns $L \propto N^2$ and

$$\frac{L_1}{L_2} = \frac{N_1^2}{N_2^2}$$

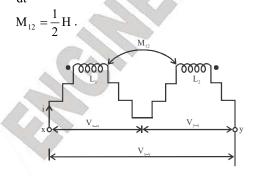
$$L_2 = \left(\frac{N_2}{N_1}\right)^2 \times L_1$$

$$= \left(\frac{1200}{600}\right)^2 \times 0.075 = 0.3H$$

$$M = \frac{N_2\phi_{12}}{L_1} = \frac{1200 \times 0.4 \times 10^{-3}}{4} = 0.12H$$

$$k = \frac{\phi_{12}}{\phi_1} = \frac{0.4}{0.5} = 0.8$$
Alternatively, $k = \frac{M}{\sqrt{L_1}} = \frac{0.12}{\sqrt{0.075 \times 0.3}} = 0.8$

Example 2. With reference to fig. Find v_{x-y} , if $\frac{di}{dt} = 300 \text{ A/sec}$. Assume $L_1 = L_2 = 1 \text{ H}$ and



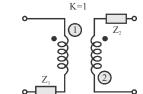
Applying KVL in the respective loops,

$$\mathbf{v}_{\mathbf{x}-\mathbf{n}} = \mathbf{L}_1 \frac{\mathrm{di}}{\mathrm{dt}} - \mathbf{M}_{12} \frac{\mathrm{di}}{\mathrm{dt}}$$

[As the coils are electrically in series, hence, the same current I passes through the coils; M is negative]

Or
$$v_{x-n} = \frac{di}{dt} - 0.5 \frac{di}{dt} = 0.5 \frac{di}{dt}$$
 ... (i)
Similarly, $v_{y-n} = L_2 \frac{di}{dt} - M_{12} \frac{di}{dt}$
 $= 1. \frac{di}{dt} - 0.5 \frac{di}{dt} = 0.5 \frac{d}{dt}$... (ii)
 $\therefore v_{x-y} = v_{x-n} + v_{y-n}$
 $= 0.5 \frac{di}{dt} + 0.5 \frac{di}{dt} = \frac{di}{dt}$
i.e., $|v_{x-y}| = 300V$

Example 3. Two impedances Z_1 and Z_2 are connected in series with the primary and secondary winding of an ideal transformer where the primary coil has $J2\Omega$ and the secondary coil has $j3\Omega$ reactance. Find the mutual reactance and inductance if $\omega = 100$ rad/sec.



Solution. In ideal transformer, K = 1 $\therefore M = \sqrt{L_1 L_2}$ i...e, $X_M = \sqrt{X_1 X_2}$ X_1 is reactance of primary coil X_2 is reactance of secondary coil X_M is mutual reactance i.e. $X_M = \sqrt{j2 \times j3} = j2.45\Omega$ but $X = 2\pi fL$

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CHAPTER - 5 NETWORK THEOREMS

5.1 THEVENIN'S THEORM

Any two terminal bilateral linear circuit can be replaced by an equivalent circuit consisting of a voltage source and a series resistor.

5.1.1 Steps for Solving a Network using Thevenin's Theorem

1. Remove the load resistor (R_L) and find the open circuit voltage (V_{oc}) across the open circuited load terminals.

2. Find the Thevenin's resistance (R_{TH})

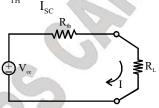
1. If Circuit contains only Independent Sources

Deactivate the constant sources (for voltage source remove it by short circuit and for the current source remove it by open circuit) and find the internal resistance (R_{TH}) of the source side looking through the open circuited load terminals.

2. For the circuits containing dependent sources in addition to or in absence of independent sources

Find V_{OC} by open circuiting the load terminals. Then short the load terminals and find the short circuit current (I_{SC}) through the shorted terminals.

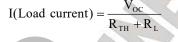
The venin's resistance is given as: $R_{TH} = \frac{V_{OC}}{I_{SC}}$



(i) Obtain the venin's equivalent circuit by placing $R_{\rm TH}$ in series with $V_{\rm OC}$

(ii) Reconnect R_L across the load terminals.

5.1.2 Thevenin's Equivalent Network



If only dependent sources are present in circuit, $R_{Th} = \frac{V_{test}}{1_{test}}$; $I_{test} = 1A$

 V_{test} is calculated across the load by short circuiting it, and current of 1A flows through the short circuited branch as I_{test} . Then $R_{TH} = V_{test}$

Dividing equation (ii) by equation (i)

$$3 = \sqrt{\frac{624 + C_s}{60.4 + C_s}}$$

$$\Rightarrow$$
 543.6+9C_s = 624+C_s

$$\Rightarrow$$
 8C_s = 80.4

$$\Rightarrow$$
 C_s = 10.05 pF

Sol. 9. (a)

In an L-C function,

(i) Poles and zeros are alternate on $j\boldsymbol{\omega}$ axis.

(ii) There is either a pole or a zero at origin and infinity.

(iii) The highest and lowest powers of s in numerator and denomenator can differ at the most by 1.

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CHAPTER - 7

LAPLACE TRANSFORMATION AND ITS APPLICATION IN CIRCUIT ANALYSIS

7.1 LAPLACE TRANSFORMATION

The Laplace transformation of a function f(t) is defined as

$$F(s) = Lf(t) = \int_{0}^{\infty} f(t)e^{-st}dt$$

Where as in complex frequency s being the intermediate or transformation variable.

7.1.1 Laplace Transform of a Derivative $\left[\frac{df(t)}{dt}\right]$

Lf'(t) = sF(s) - f(0+)

7.1.2 Laplace Transform of an Integral $\int f(t)dt$

 $L\left(\int f(t)dt\right) = \frac{1}{s}\int f(t)dt |_{0+} + \frac{1}{s}F(s)$ $\left[\int f(t)|_{0+} \text{ gives the value of the integral at } t = 0 + \frac{1}{s}F(s)\right]$

7.1.3 Frequency Shifting $L(e^{at}f(t)) = F(s-a)$

$$L(e^{-at}f(t)) = F(s+a)$$

7.2 LAPLACE TRANSFORM OF COMMON FORCING FUNCTIONS

	<i>f</i> (t)	F(s)	<i>f</i> (t)	F(s)
	u(t)	$\frac{1}{s}$	$e^{-\alpha t}t^n$	$\frac{\underline{ n }}{(s+\alpha)^{n+1}}$
	e ^{-at}	$\frac{1}{s+\alpha}$	e ^{-at} sin@t	$\frac{\omega}{\left(s+\alpha\right)^2+\omega^2}$
2	sinwt	$\frac{\omega}{s^2 + \omega^2}$	$e^{-\alpha t} \cos \omega t$	$\frac{s+a}{(s+\alpha)^2+\omega^2}$
2	cosωt	$\frac{\omega}{s^2 + \omega^2}$	δ(t)	1
	t	$\frac{1}{s^2}$	Sinh θt	$\frac{\theta}{s^2-\theta^2}$

CHAPTER - 8 RESONANCE

8.1 RESONANCE

Resonance in electrical circuits consisting of passive and active elements represents a particular state of the circuit when the current or voltage in the circuit is maximum or minimum with respect to the magnitude of excitation at a particular frequency, the circuit impedance being either minimum of maximum at the power factor unity.

The phenomenon of resonance is observed in both series or parallel a.c. circuits comprising of R, L and C and excited by an a.c. source.

8.2 SERIES RESONANCE

$$V_{\text{in sinot}}$$
 \downarrow Z_{in} \uparrow 0

$$Z_{in} = \frac{V}{I} = R + j(\omega L - \frac{1}{\omega c})$$

For resonance V & I must be in same phase So for some frequency $\omega = \omega_0$

$$Z_{in} = \mathbf{R} + \mathbf{j}_0 \Rightarrow \omega_0 \mathbf{L} - \frac{1}{\omega_0 \mathbf{C}} = \mathbf{0} \Rightarrow \omega_0 = \frac{1}{\sqrt{\mathbf{LC}}}$$
$$\mathbf{I} = \frac{\mathbf{V}}{|\mathbf{Z}|} = \frac{\mathbf{V}}{\sqrt{\mathbf{R}^2 + (\omega \mathbf{L} - 1/\omega \mathbf{c})^2}} \text{ at } \omega_0, \ \mathbf{I} = \frac{\mathbf{V}}{\mathbf{R}}$$

$$A \downarrow B \\ Q_3 \\ Q_3 \\ Q_1 \\ Q_1 \\ Q_1 \\ Q_2 \\ Q_3 \\ Q_2 > Q_1$$

Points A & B are half power or 3dB points because $20 \log_{10} \left(\frac{1}{2}\right) = 3 dB$

Band width of circuit $\Delta \omega = BW = \omega_2 - \omega_1$ Quality factor

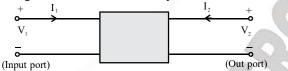
$$Q_0 = 2\pi \left[\frac{\text{Max energy stored}}{\text{Total energy last per perior}} \\ Q_0 = 2\pi \left[\frac{\omega_L + \omega_C}{P_R T} \right] \right]$$



CHAPTER - 9 TWO PORT NETWORKS

9.1 INTRODUCTION

The terminal pair is called as a "port". If the current entering one terminal of a pair is equal and opposite to the current leaving the other terminal of the pair.



9.2 TWO-PORT NETWORK

A two-port network is shown, by which we observe that a two-port network is represented by a black box with four variables, namely, two voltages (V_1, V_2) and two currents (I_1, I_2) which are available for measurements and are relevant for the analysis of two port networks. Of these four variables which two variable may be considered `independent` and which two `dependent` is generally decided by the probable under consideration

Two Port Parameters						
Name	Express	In terms of	Matrix Equation			
Open circuit impedance [Z]	V ₁ , V ₂	I ₁ ,I ₂	$\begin{bmatrix} \mathbf{V}_1 \\ \mathbf{V}_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{I}_2 \end{bmatrix}$			
Short-circuit admittance [Y]	I ₁ , I ₂	V ₁ , V ₂	$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$			
TransmissionorChain[T]or[ABCD]	V ₁ , I ₂	V ₂ ,I ₂	$\begin{bmatrix} \mathbf{V}_1 \\ \mathbf{I}_1 \end{bmatrix} = \begin{bmatrix} \mathbf{A} \ \mathbf{B} \\ \mathbf{C} \ \mathbf{D} \end{bmatrix} \begin{bmatrix} \mathbf{V}_2 \\ -\mathbf{I}_2 \end{bmatrix}$			
Inverse Transmission [T']	V ₂ ,I ₂	$V_1, -I_1$	$\begin{bmatrix} \mathbf{V}_2 \\ \mathbf{I}_2 \end{bmatrix} = \begin{bmatrix} \mathbf{A}'\mathbf{B}' \\ \mathbf{C}'\mathbf{D}' \end{bmatrix} \begin{bmatrix} \mathbf{V}_1 \\ -\mathbf{I}_1 \end{bmatrix}$			
Hybrid (h)	V ₁ , I ₂	I_1, V_2	$\begin{bmatrix} \mathbf{V}_1 \\ \mathbf{I}_2 \end{bmatrix} = \begin{bmatrix} \mathbf{h}_{11} \ \mathbf{h}_{12} \\ \mathbf{h}_{21} \ \mathbf{h}_{22} \end{bmatrix} \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{V}_2 \end{bmatrix}$			
Inverse hybrid (g)	I ₁ , V ₂	V ₁ ,I ₂	$\begin{bmatrix} I_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} g_{11} \ h_{12} \\ g_{21} \ h_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ I_2 \end{bmatrix}$			

9.3 OPEN CIRCUIT IMPEDANCE (Z) PARAMETERS

Expressing two-port voltages in terms of two-port currents $(V_1, V_2) = f(I_1, I_2)$

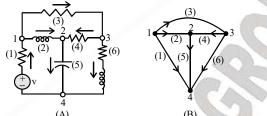
$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} \text{ or } [V] = [Z][I]$$



CHAPTER - 10 **GRAPH THEORY**

10.1 IMPORTANT DEFINITIONS

1. Graph: It is the collection of nodes and Branch of a network.



(A) (B) **2. Branch:** Each oriented line segment of the graph is called branch.

3. Node: The end point of a branch is called node.

4. Incident Branch: Branch whose end fall on a node is called incident branch.

5. Connected and Non-Connected Graph

If there exists a path between every pair of nodes of a graph, then the graph is called connected graph, otherwise graph is called non-connected graph.

6. Degree of Node: Degree of Node is the number of branches incident on the node.

7. Subgraph: A portion of graph is called subgraph

8. Path: Path is transverse from one node to another node

9. Loop: Loop is a collection of branches in a graph which form a closed path.

10. Tree: The collection of minimum no. of branches connecting all the nodes of a graph without making a loop.

A single graph can have many no. of trees. The no. of trees for a given graph = n - 1where $n \rightarrow no$. of nodes

11. Twig: Branch of a tree is called a twig.

12. Cotree: Remaining part of a graph after removal of twigs is called cotree. It is collection of links.

13. Links: are the branches removed from the graph to make a tree. Total no. of branch of a graph are given by b = (n-1) + Ln is no. of nodes L is No. of links No. of twigs = (n-1) = no. of KCL equation



CHAPTER - 11 NETWORK FUNCTIONS

11.1 INTRODUCTION

The basic definition of one port and two port network being discussed earlier, here we will discuss about the transform of excitation and response along with their relations. A network function exhibits the relationship between the transform of the source or excitation to the transform of the response for a electrical network. Further to this, we will discuss the stability of the network function mathematically formulating the network function mathematically formulating the network function through "pole-zero" concept.

11.2 DRIVING POINT IMPEDANCE AND ADMITTANCE

The driving point impedance of a one port network is defined as

$$Z(s) = \frac{V(s)}{I(s)}$$

While the driving point admittance is given as

$$Y(s) = \frac{I(s)}{V(s)}$$

For the one port network

Similarly, for the two port network, the driving point impedance and admittance at port 1 is defined as

$$Z_{11}(s) = \frac{V_1(s)}{I_1(s)} \qquad \dots (iii)$$

and $Y_{11}(s) = \frac{I_1(s)}{V_1(s)} \qquad \dots (iv)$

While the driving point impedance and admittance at the port 2 are designated as

$$Z_{22}(s) = \frac{V_2(s)}{I_2(s)}$$

and $Y_{22}(s) = \frac{I_2(s)}{V_2(s)}$

11.3 TRANSFER IMPEDANCE AND ADMITTANCE

Transfer impedance is defined as the ratio of transform voltage at output port to the transformed current at the input port of a two port network.

This gives,
$$Z_{12}(s) = \frac{V_2(s)}{I_1(s)}$$

In a similar way, the transfer admittance is defined as the ratio of current transform at output port to the voltage transform at the input port. It is given as

$$Y_{12}(s) = \frac{I_2(s)}{V_1(s)}$$

...(vii)

... (vi)

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... (i)

... (ii)

...(iv(a))

... (v)

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